



# Software-Defined Radio System for Tracking Application

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# Overview

Background

Proposed solution

Used signal processing

Simulation in GNU Radio

Measurement Results

Conclusion and Outlooks

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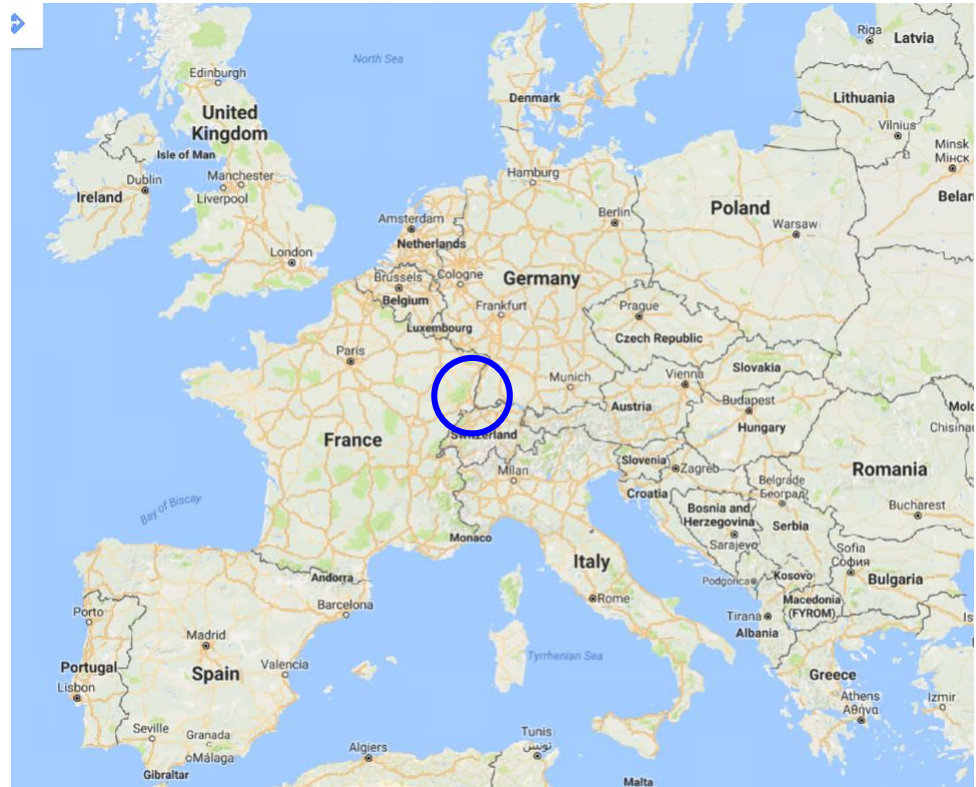
ISL: French-German Research  
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# Background

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Institute of Saint-Louis in  
France

STC group in charge of telemetry  
and bi-directional links with  
flying projectiles





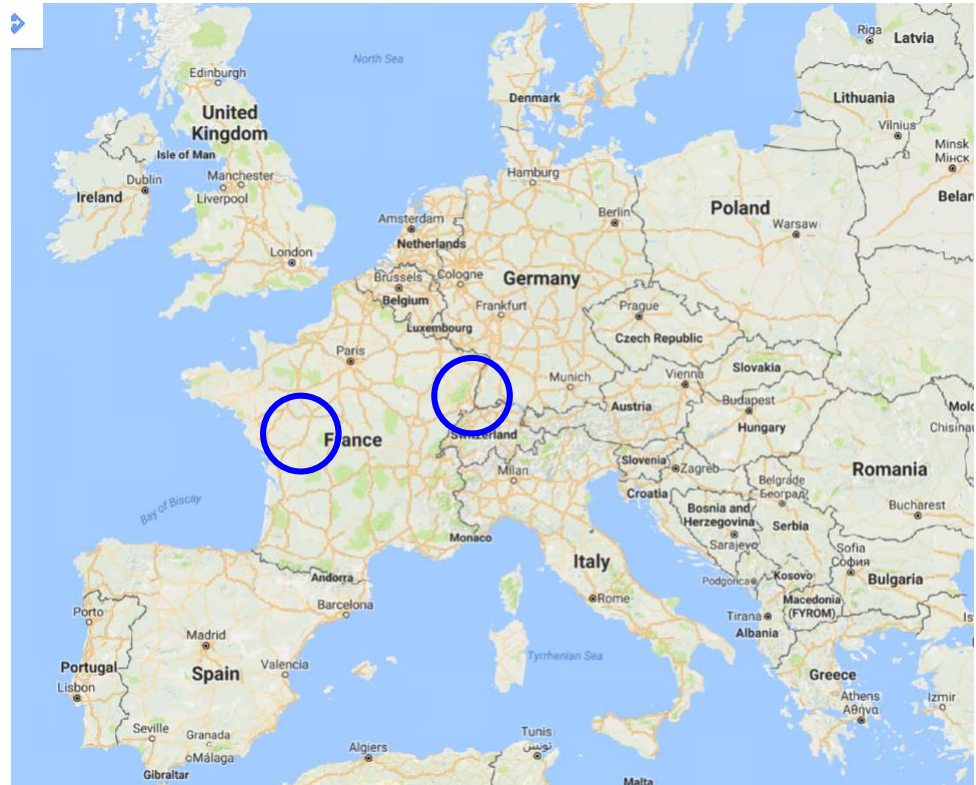
# Background

ISL: French-German Research Institute of Saint-Louis in France

STC group in charge of telemetry and bi-directional links with flying projectiles

Work developed within a PhD funded by ISL, supervised by ISL and XLIM

XLIM: A multidisciplinary Research Institute located on several geographical sites, mainly in Limoges but also in Poitiers



# Background

Many Software Defined Radio (SDR) tracking applications in research focus on tracking mobile phones, vehicles, satellites, etc [1-2] with various difficulties to overcome (indoor localization, multi-path).

[1]: V. Nambiar et al., "SDR based indoor localization using ambient WiFi and GSM signals," *2017 ICNC*.

[2]: V. Y. Vu and A. B. Delai, "Digital Solution for inter-vehicle localisation system by means of Direction-Of-Arrival," *2006 ISISPC*

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Our project:

**develop a passive high-speed projectile tracking system based on commercial SDR and antenna arrays**

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# Overview

Background

**Proposed solution**

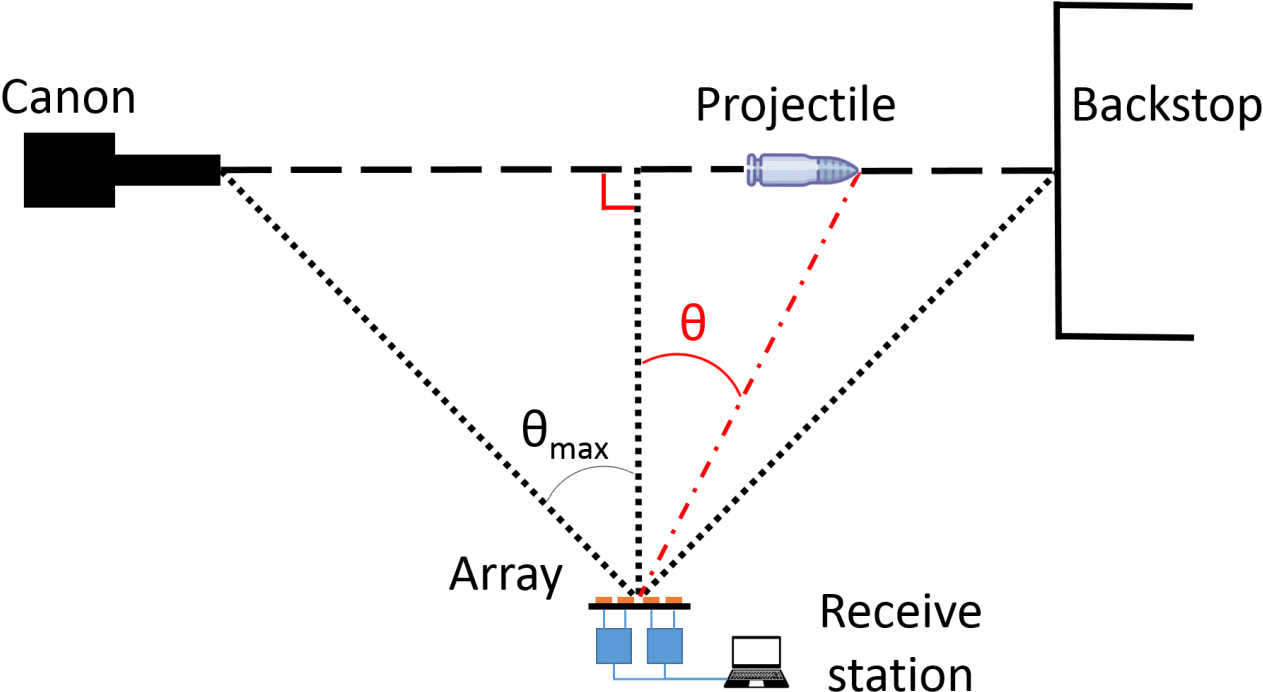
Used signal processing

Simulation in GNU Radio

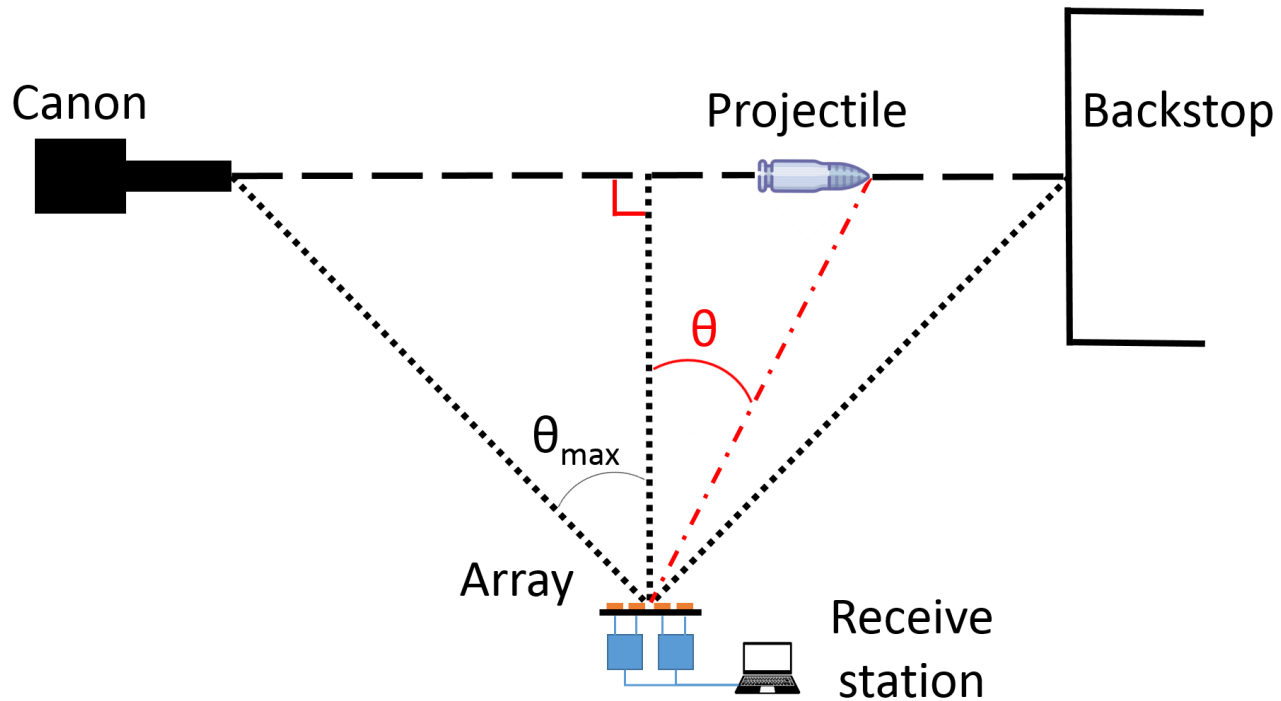
Measurement Results

Conclusion and Outlooks

# Proposed solution

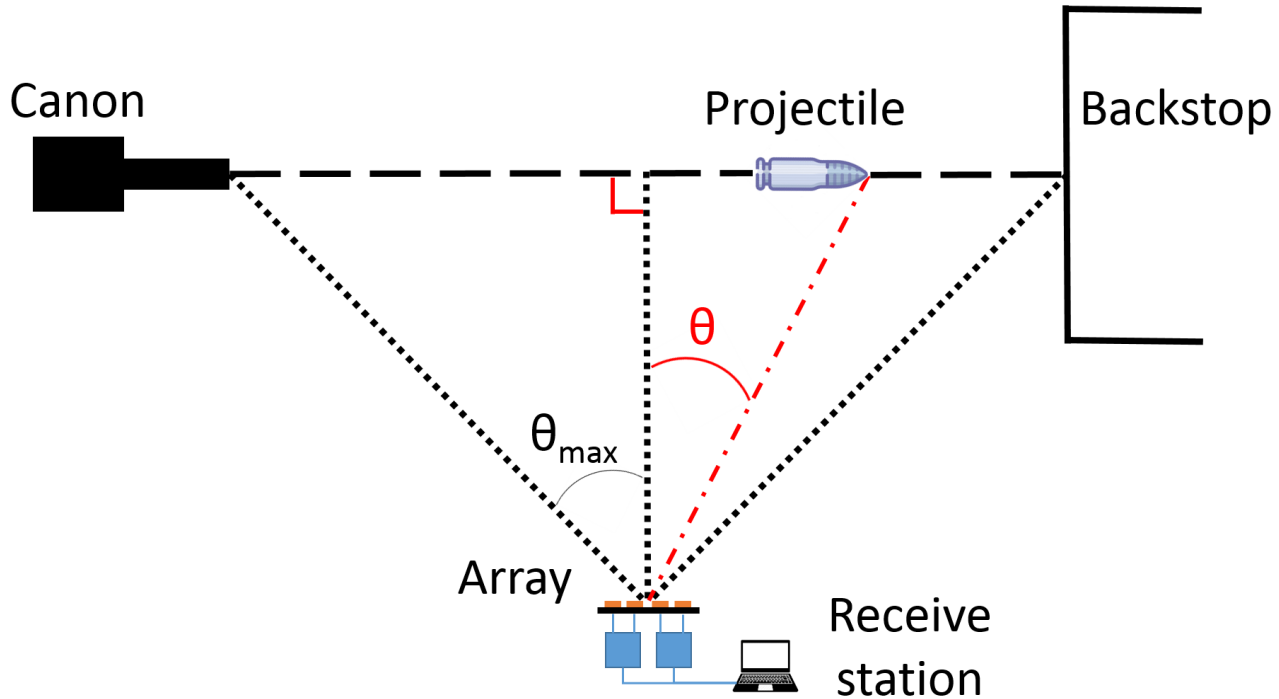


# Proposed solution



Steered antenna array electronically follows projectile (no mechanical displacement)

# Proposed solution

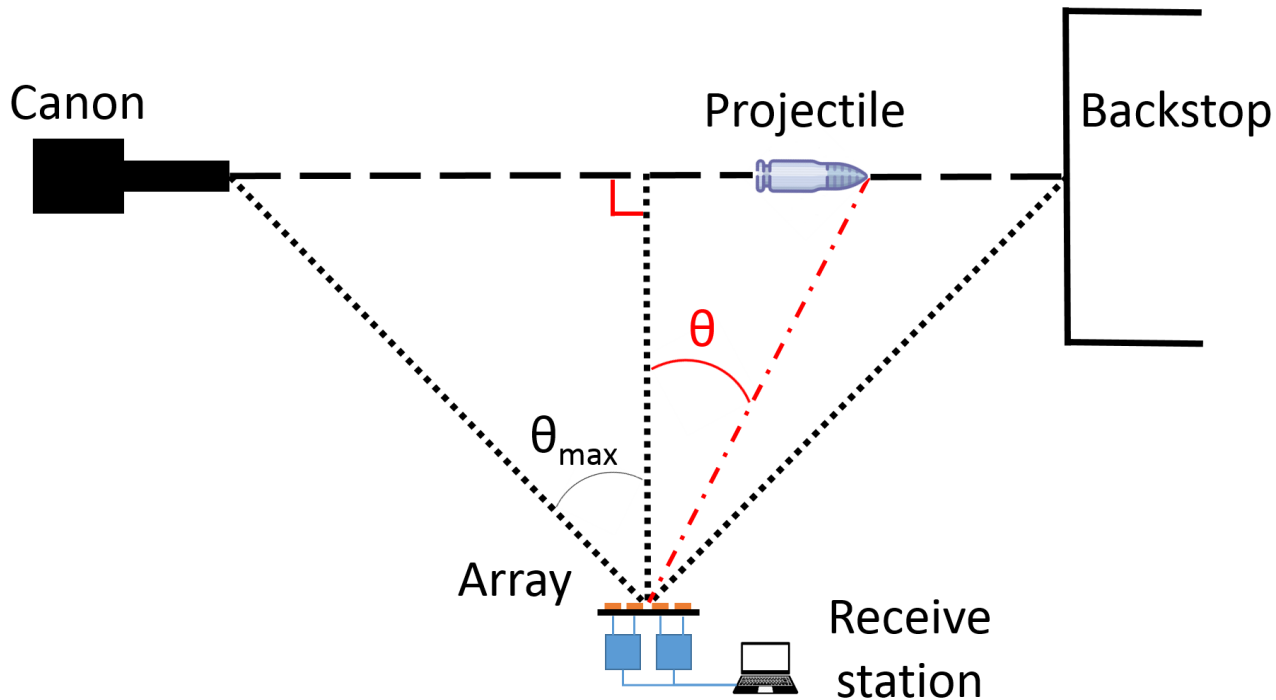


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Direction of the projectile computed from received signals (DOA)



# Proposed solution



Steered antenna array electronically follows projectile (no mechanical displacement)

Direction of the projectile computed from received signals (DOA)

Array main lobe steered towards estimated direction of the transmitter

# Overview

Background

Proposed solution

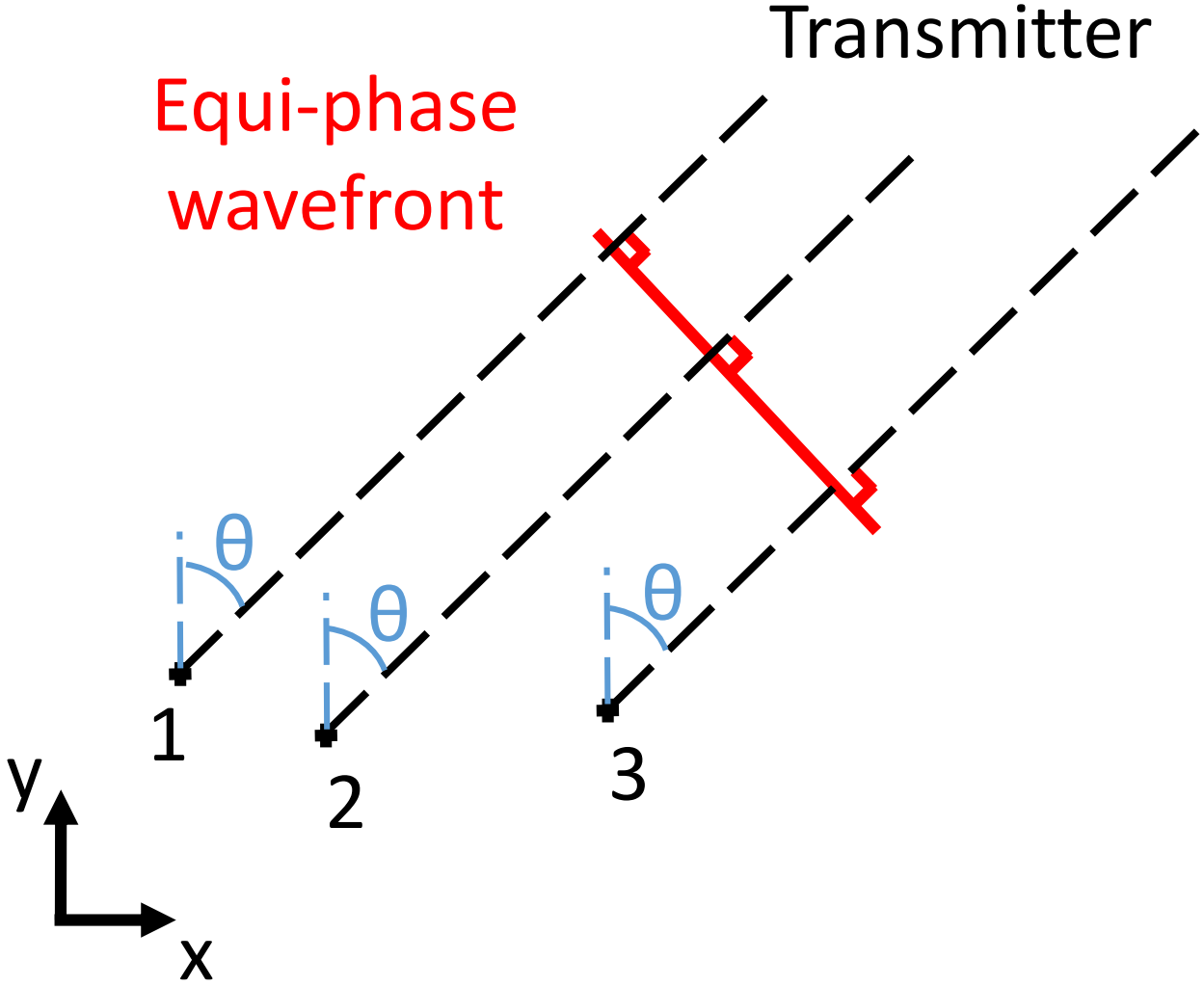
**Used signal processing**

Simulation in GNU Radio

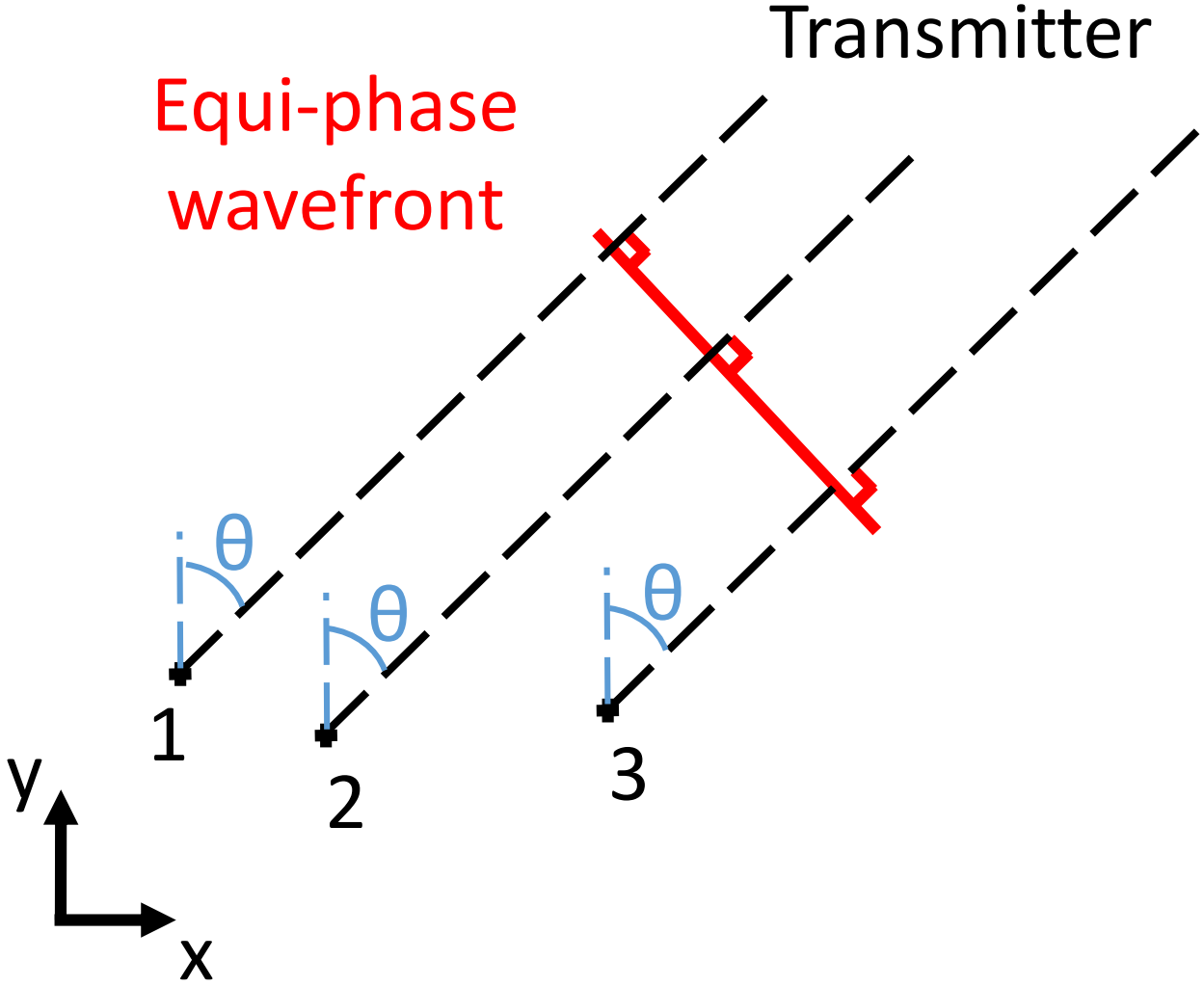
Measurement Results

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# Used signal processing



# Used signal processing



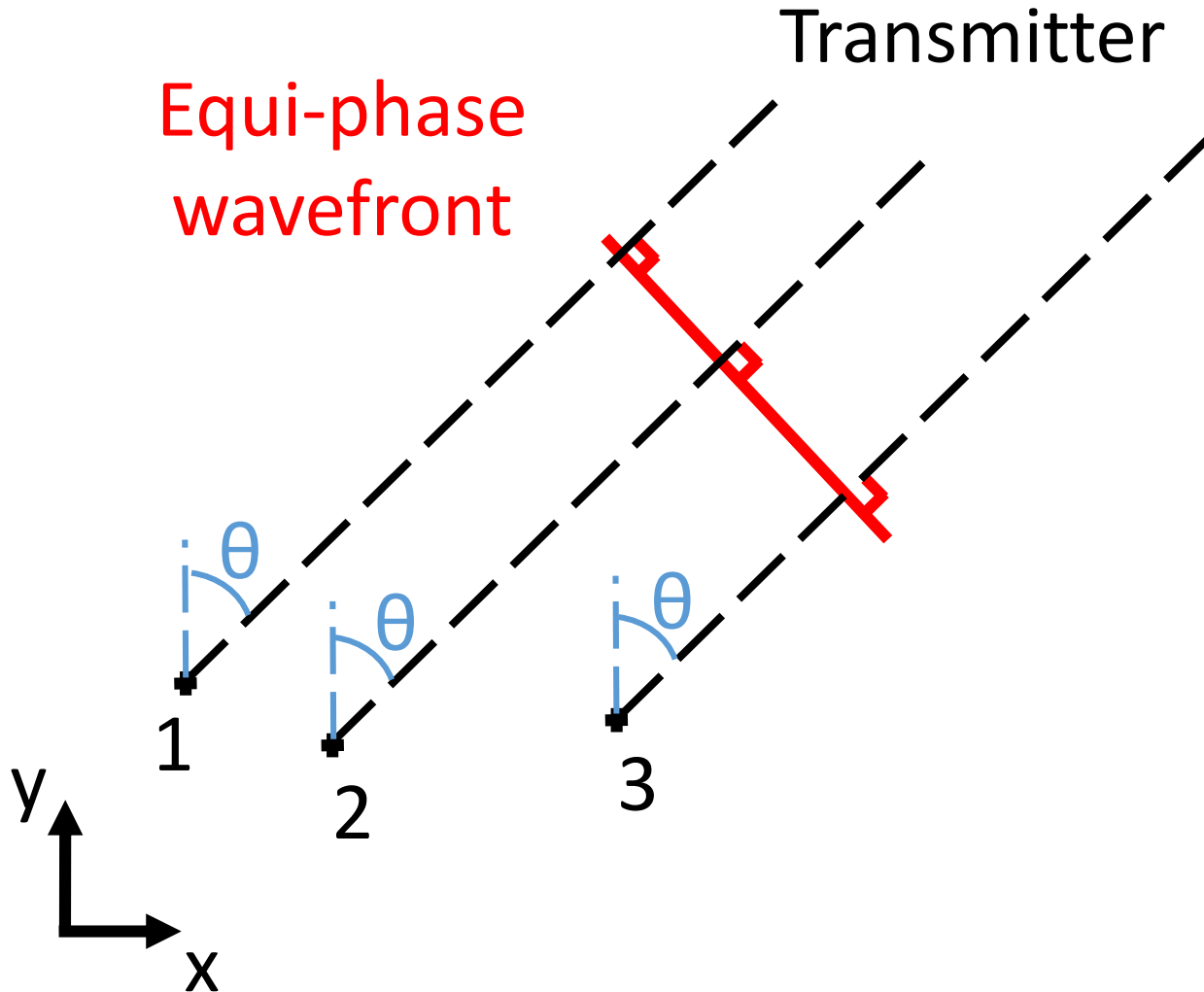
Transmitter

Equi-phase  
wavefront

Transmitted signal:

$$s_{TX} = e^{j\omega t}$$

# Used signal processing



Transmitted signal:

$$s_{TX} = e^{j\omega t}$$

Received signals:

$$s_1 = e^{j(\omega t + \varphi_1)}$$

$$s_2 = e^{j(\omega t + \varphi_2)}$$

$$s_3 = e^{j(\omega t + \varphi_3)}$$

## Used signal processing

By influencing on the phase shifts between signals:

favor constructive interference for chosen particular  $\theta$  (DOA).

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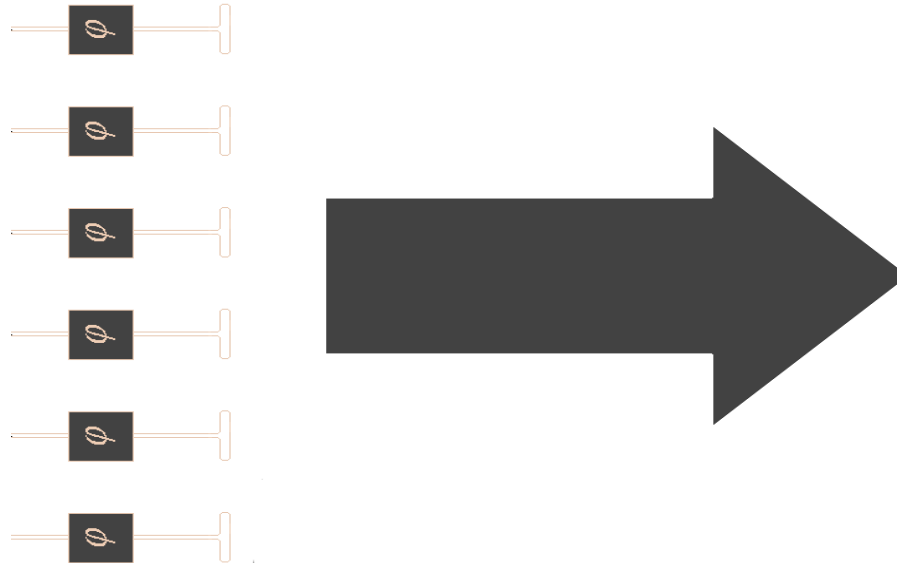


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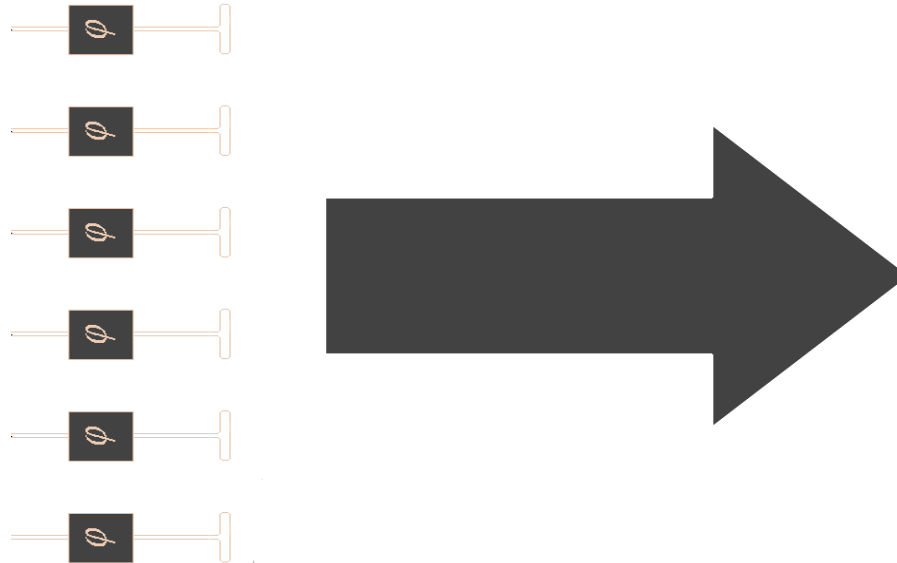


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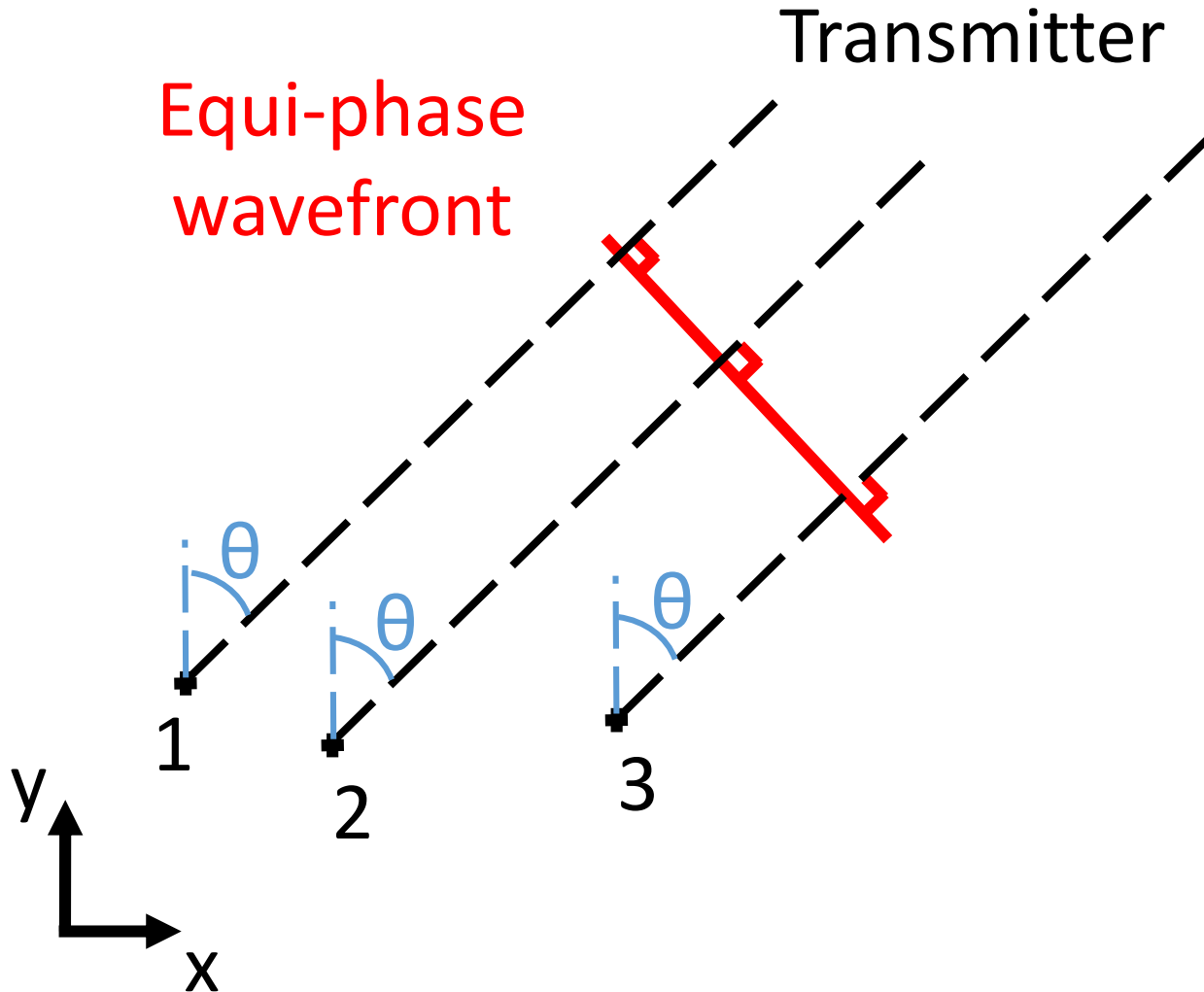
For a Uniform Linear Array (ULA), phase shifts between adjoining elements  $\varphi$ :

$$\varphi = \frac{2\pi}{\lambda} d \sin\theta$$

$\lambda$ : wavelength

$d$ : inter-element spacing

# Used signal processing



Transmitted signal:

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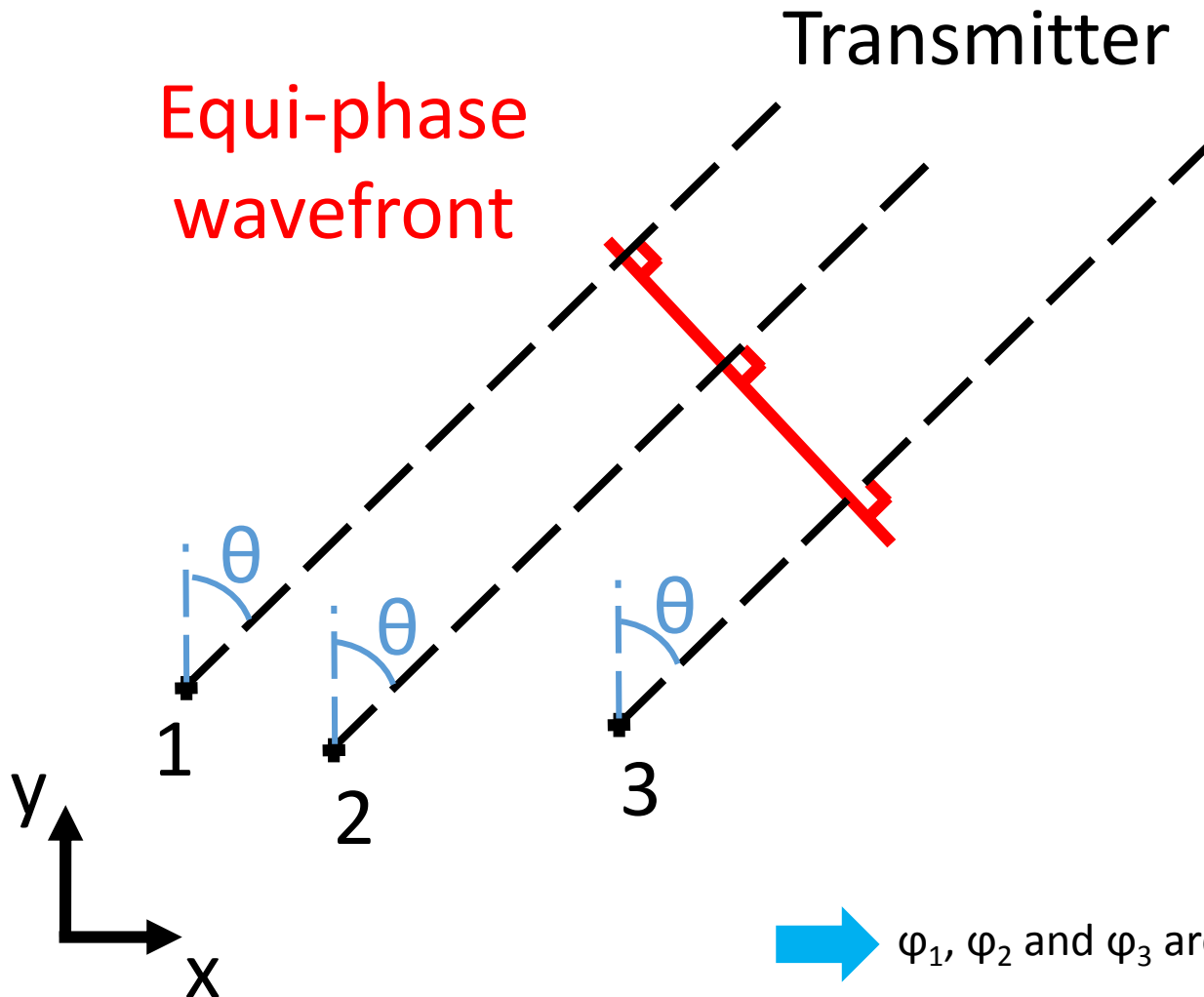
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Transmitted signal:

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Received signals:

$$s_1 = e^{j(\omega t + \varphi_1)}$$

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➡  $\varphi_1, \varphi_2$  and  $\varphi_3$  are used to estimate  $\theta$  (DOA)

# Used signal processing

DOA computed using conventional beamformer algorithm (Bartlett)

- Only one transmitting source

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The DOA is computed using the conventional beamformer algorithm (Bartlett)

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- Easy and fast to implement

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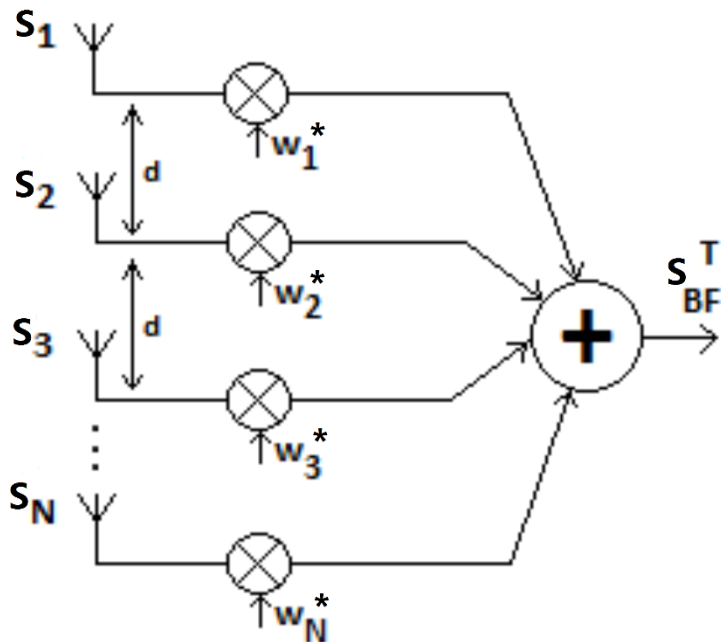
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- Less computationally demanding than MUSIC



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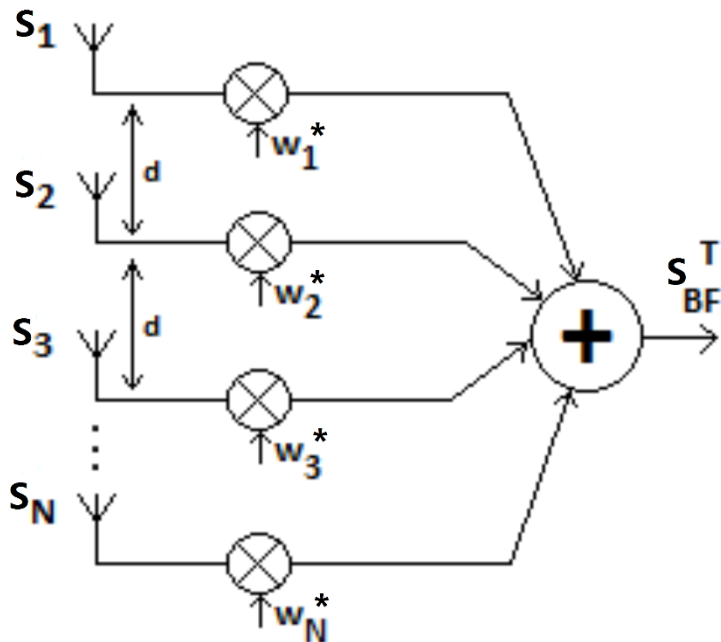
$$S = \begin{pmatrix} S_1 \\ \dots \\ S_N \end{pmatrix}$$

$$W = \begin{pmatrix} W_1 \\ \dots \\ W_N \end{pmatrix}$$

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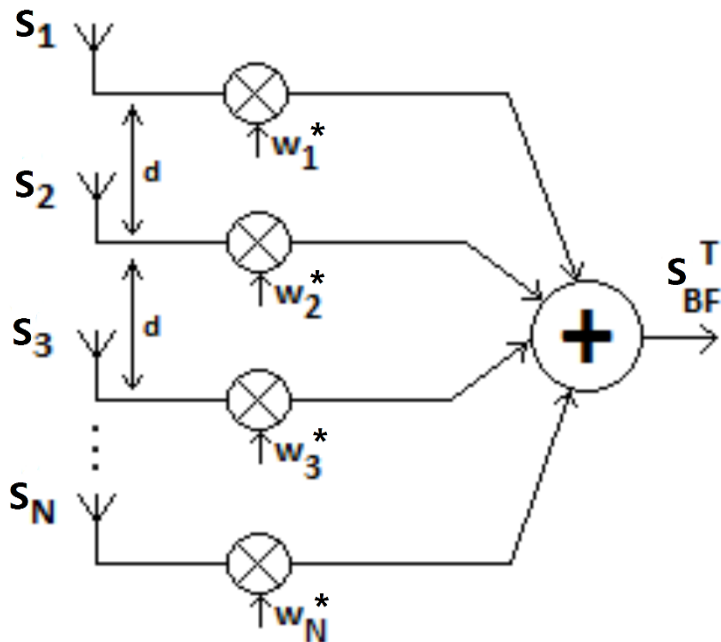
$$s = \begin{pmatrix} s_1 \\ \dots \\ s_N \end{pmatrix} \quad w = \begin{pmatrix} w_1 \\ \dots \\ w_N \end{pmatrix}$$

$$s_{BF}^T = w_1^* s_1 + \dots + w_N^* s_N = w^H s$$

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➡ Optimize  $w$  in order to maximize  $s_{BF}^T$

# Used signal processing

Work presented here focuses on reception mode.

Commercial UBX-160 & Octoclock



frequency and sampling time synchronization between channels





NO phase synchronization between channels


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 Initialization position referred as “ $\theta = 0^\circ$ ” position.

# Used signal processing

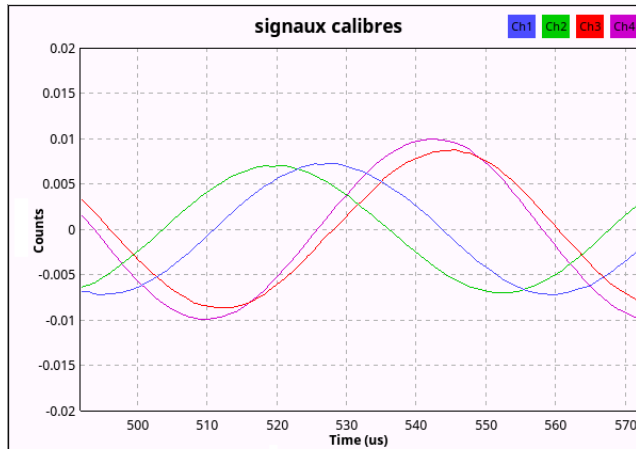
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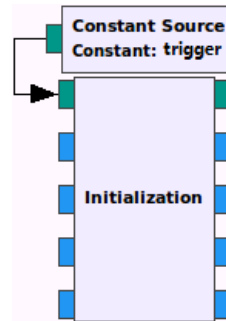
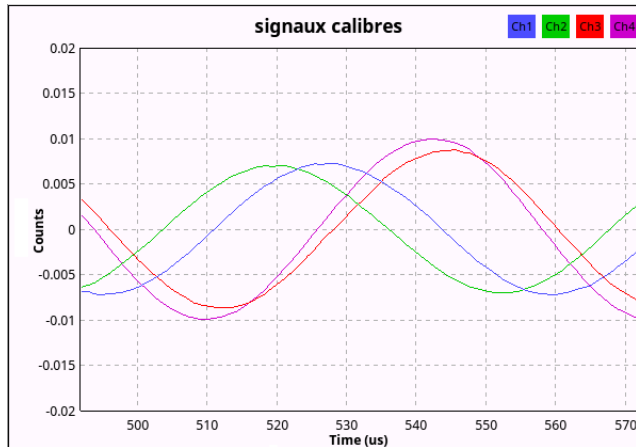
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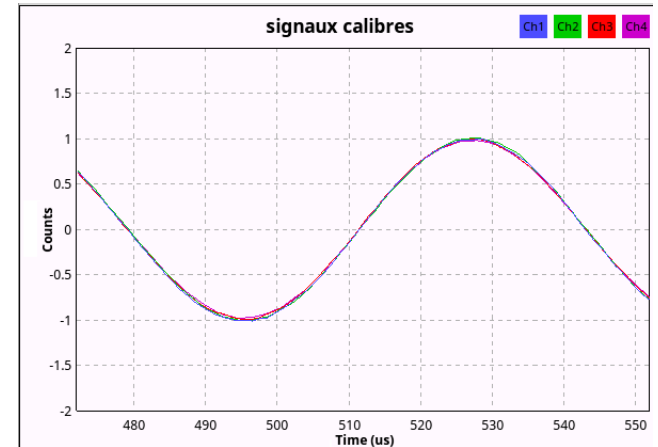
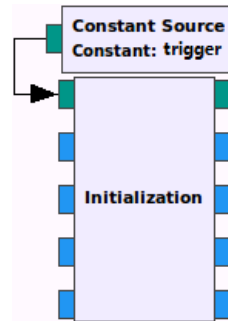
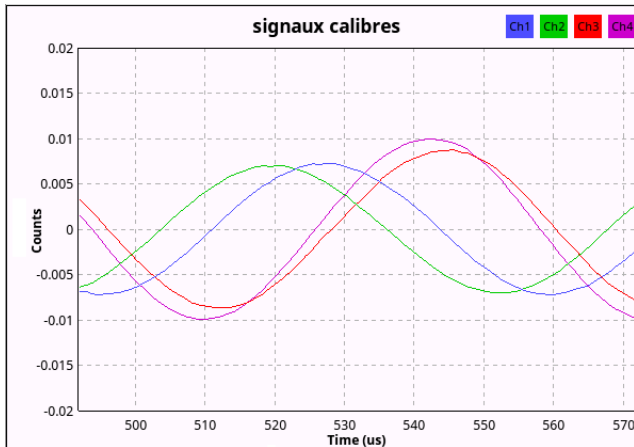
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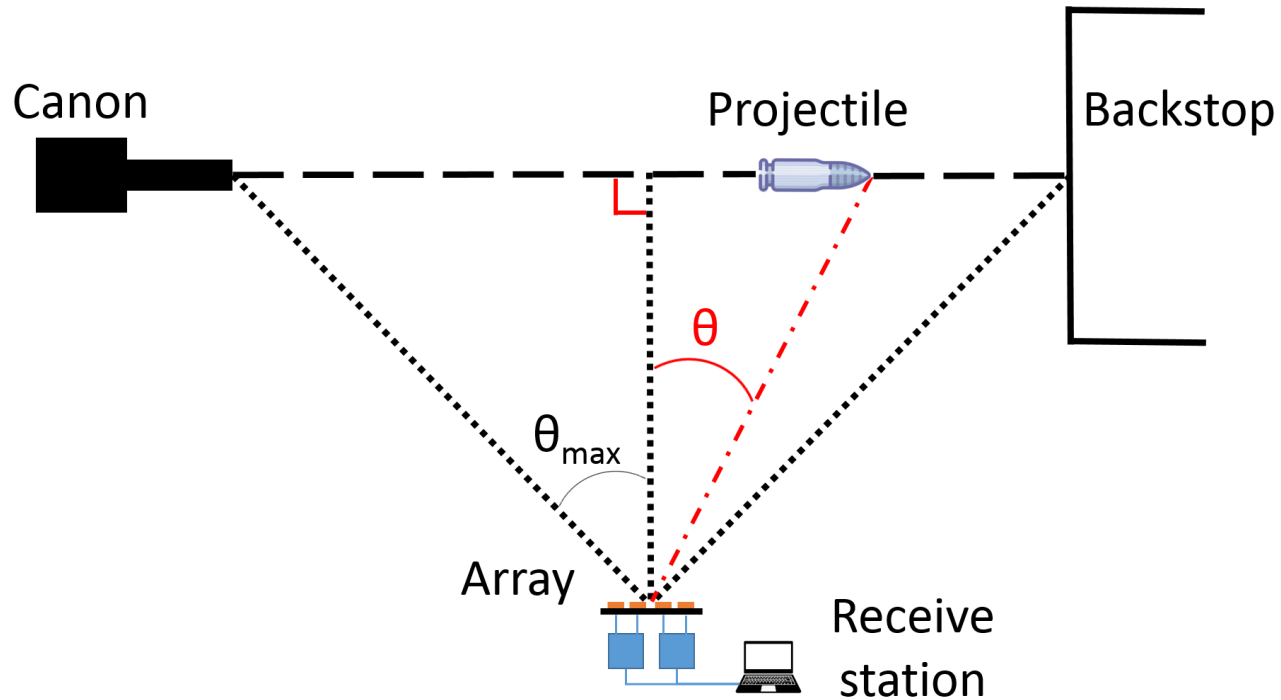
**Simulation in GNU Radio**

Measurement Results

Conclusion and Outlooks

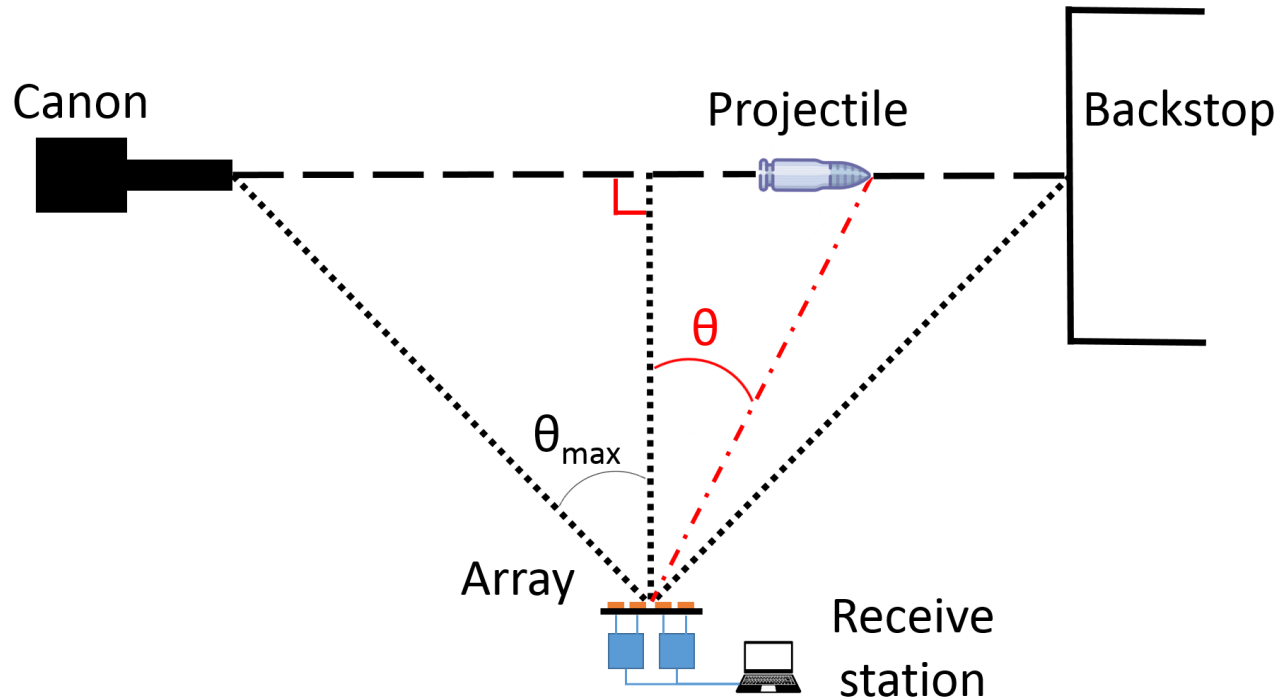
# Simulation in GNU Radio

Aimed application: passive solution for projectile following



# Simulation in GNU Radio

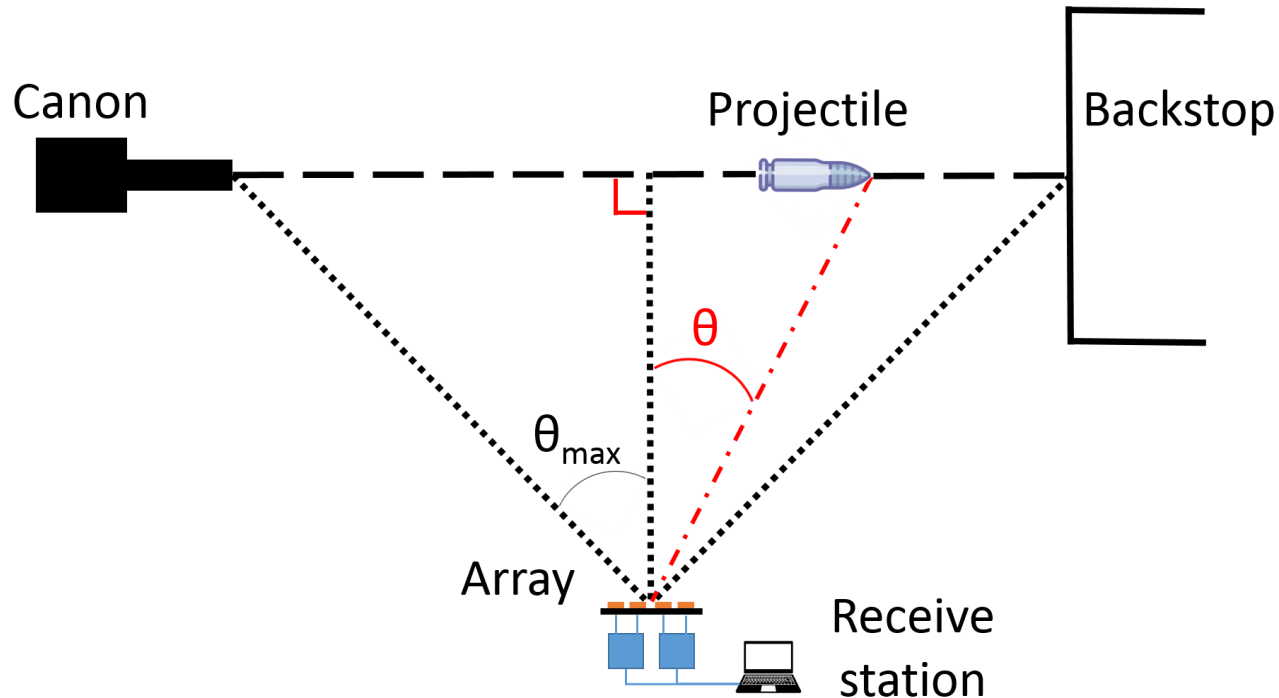
Aimed application: passive solution for projectile following



Estimated DOA used to steer array main lobe in projectile direction in real-time

# Simulation in GNU Radio

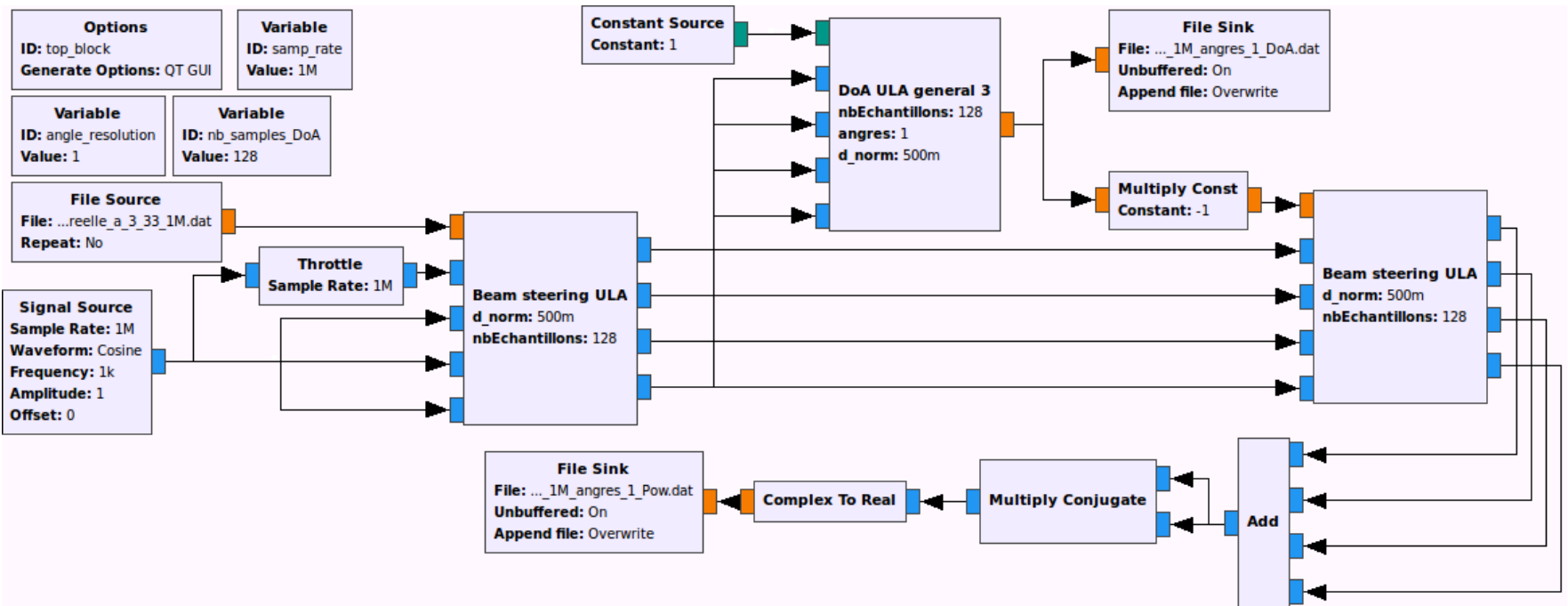
Aimed application: passive solution for projectile following



Estimated DOA used to steer array main lobe in projectile direction in real-time

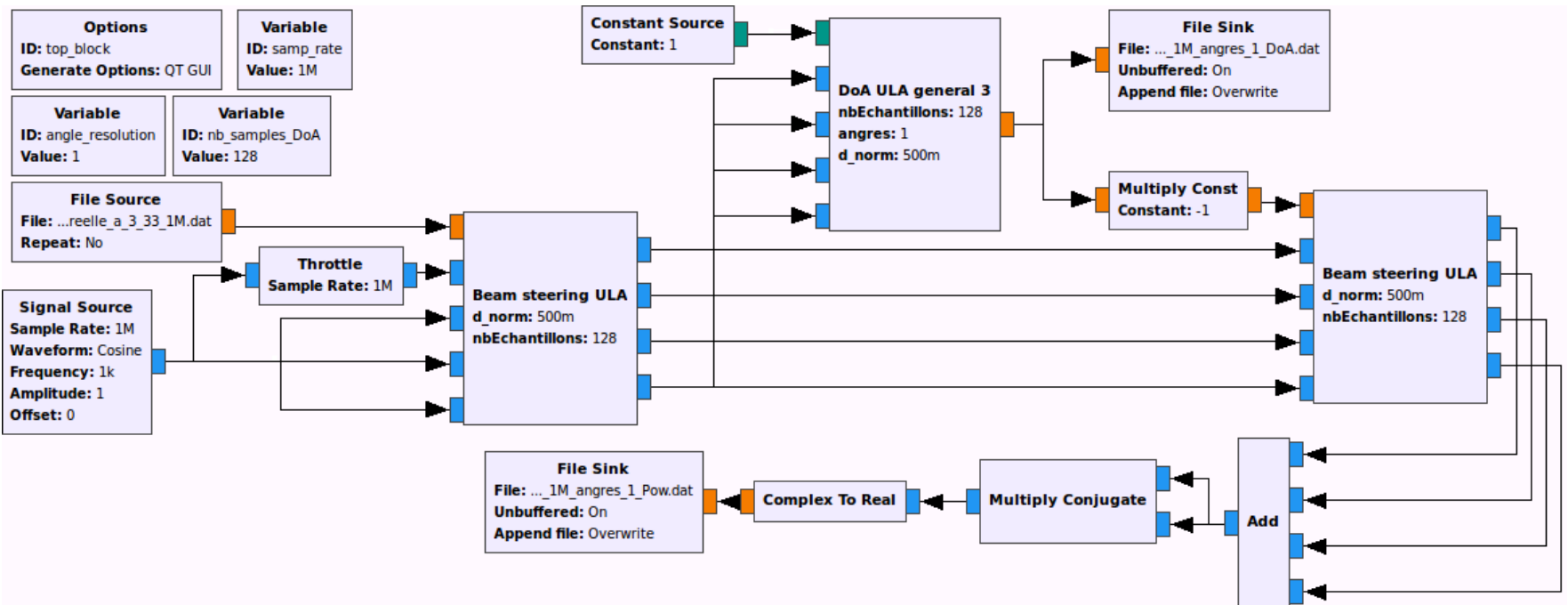
$\theta$  values were generated from a previous firing to run simulations in GNU Radio

# Simulation in GNU Radio



Expected performance for a particular setup

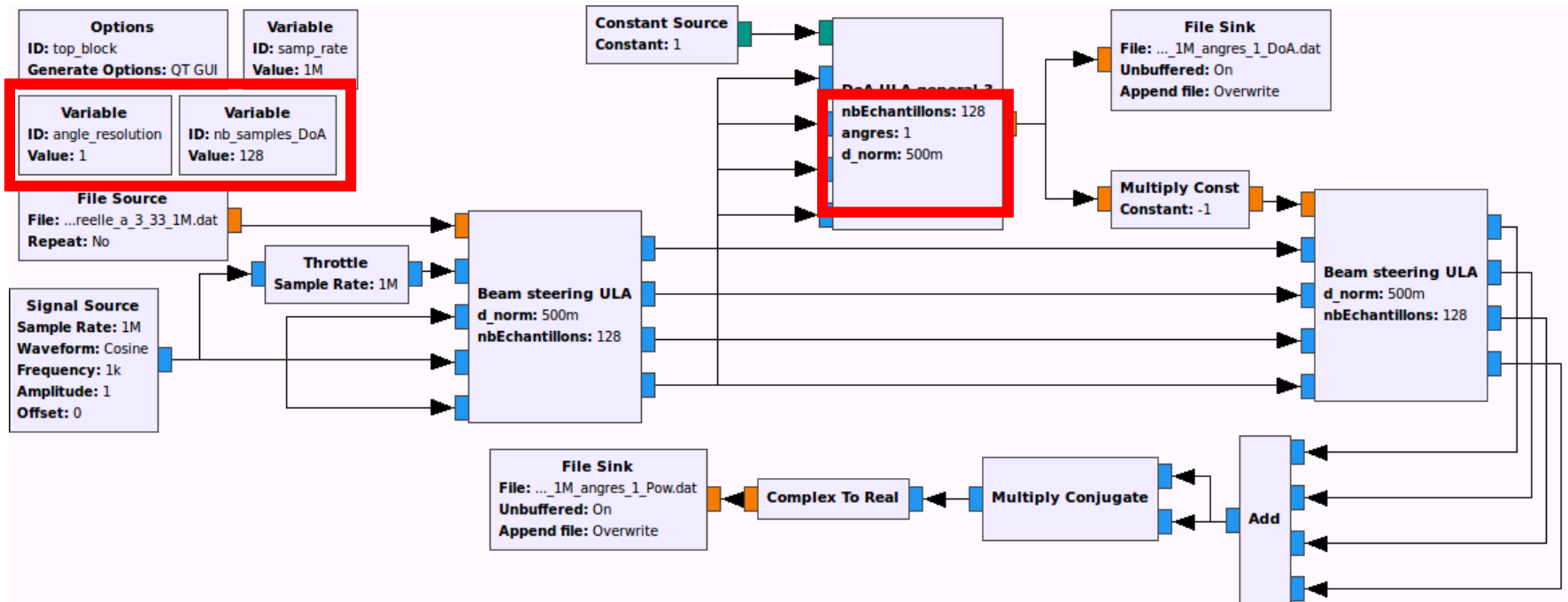
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Expected performance for a particular setup

$\theta$  values simulate sampling for projectiles of different speeds and for different sampling rates

# Simulation in GNU Radio



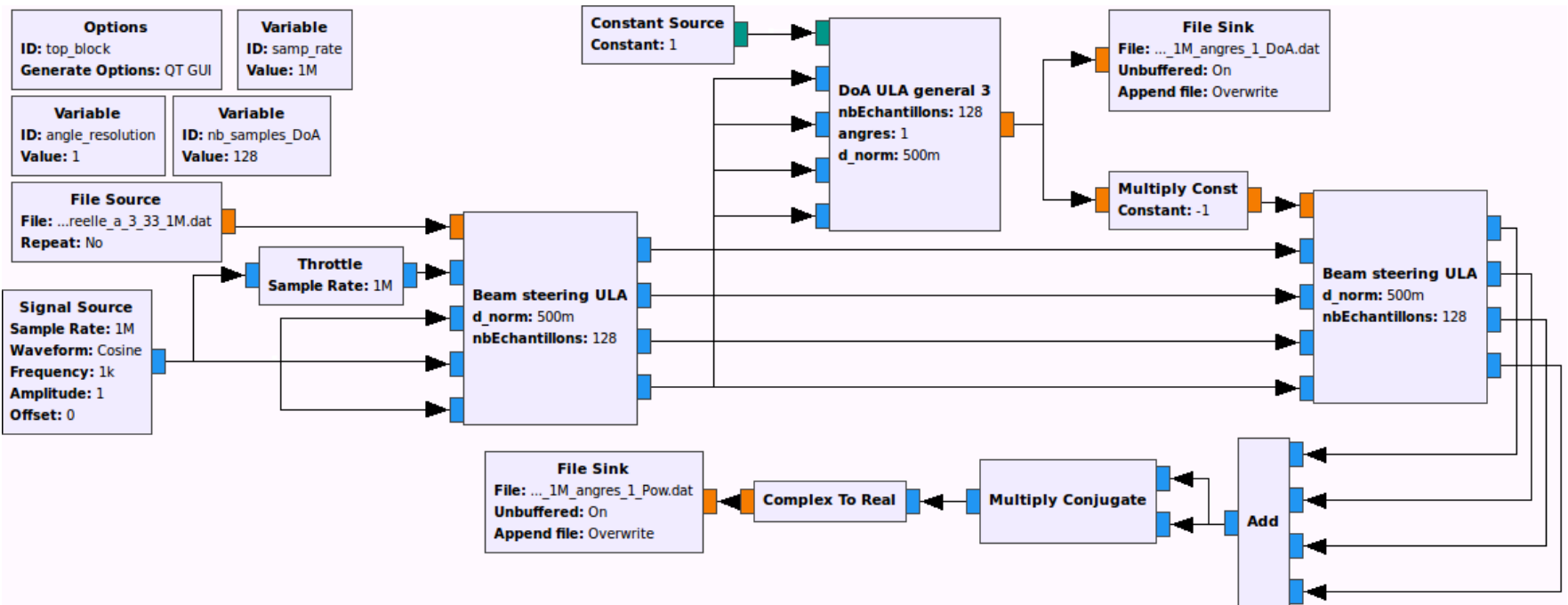
Expected performance for a particular setup

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DOA estimation performance can be adjusted by setting DOA search angular resolution and number of samples per estimation.



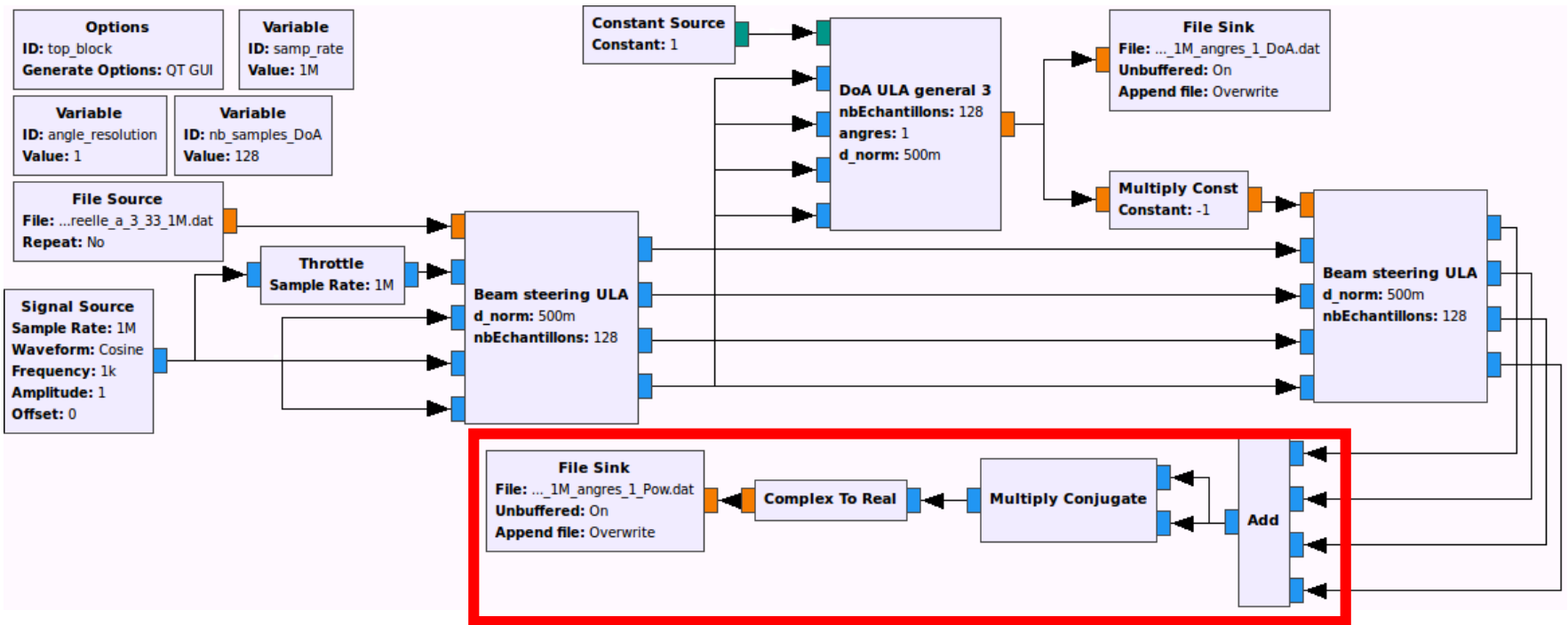
# Simulation in GNU Radio



Expected performance for a particular setup

Signal amplitudes normalized to unit

# Simulation in GNU Radio

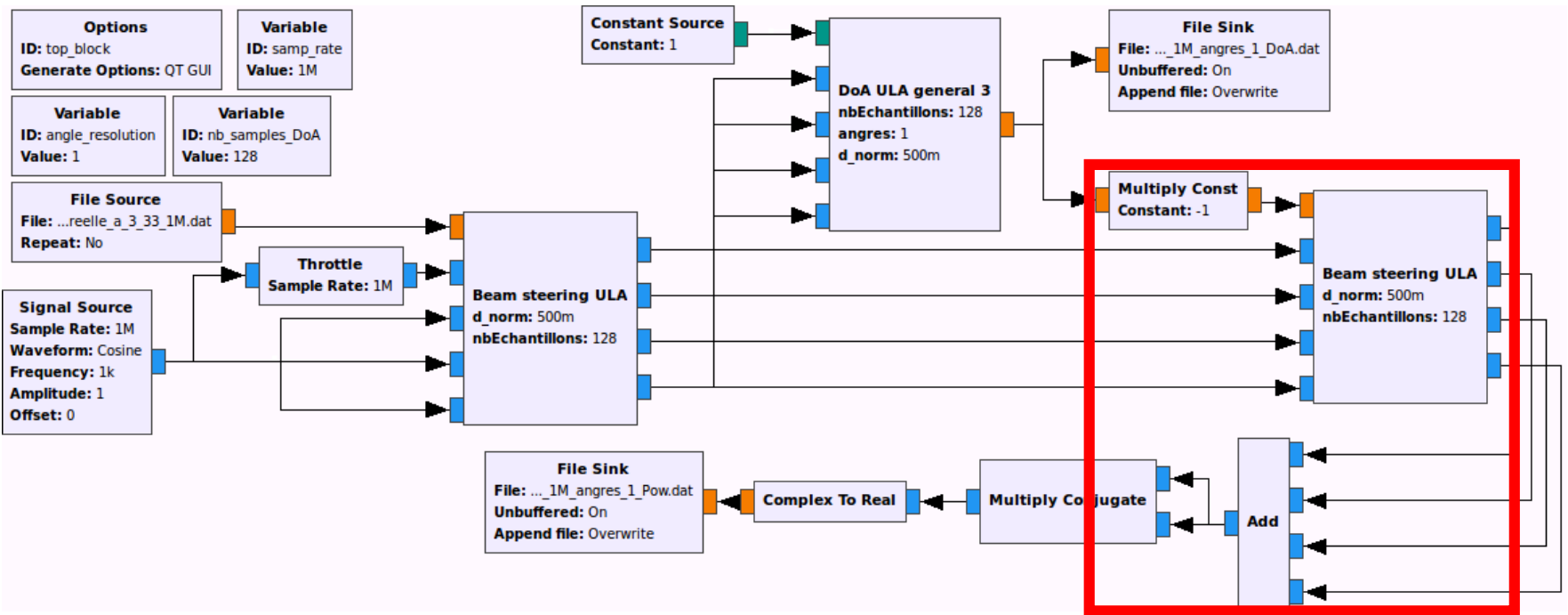


Expected performance for a particular setup

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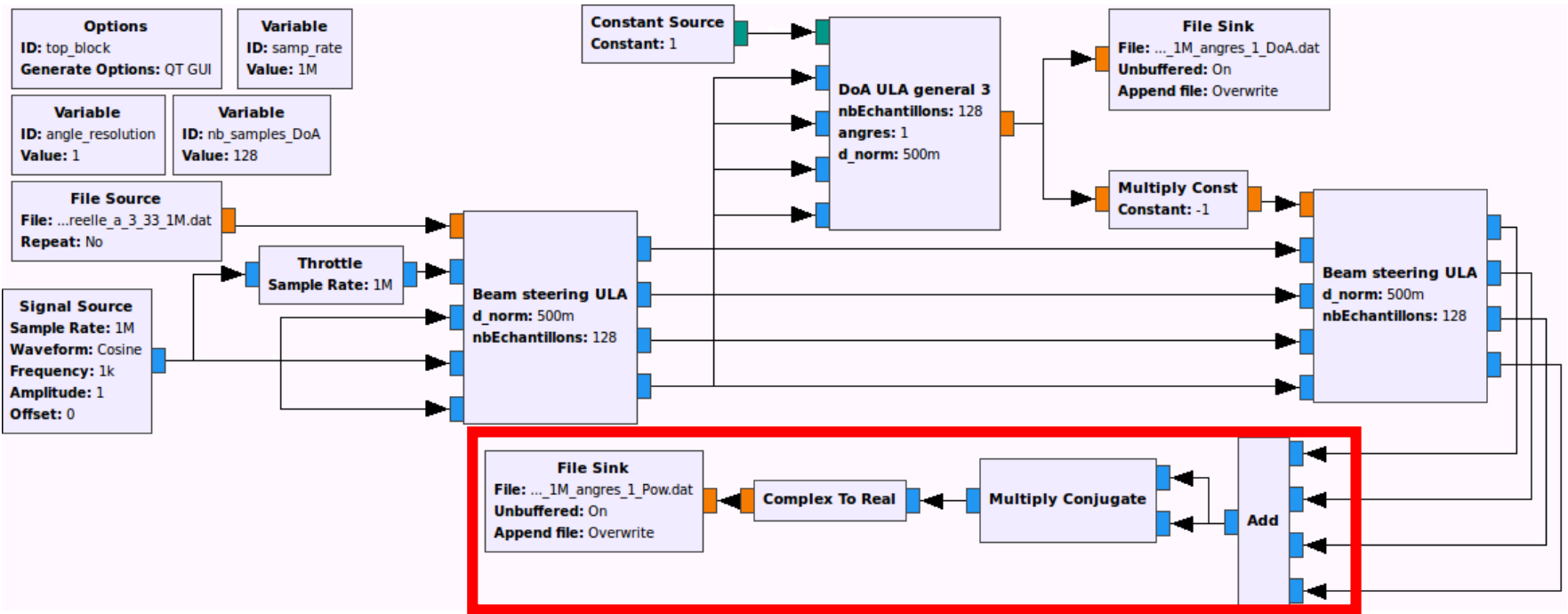
Program ability to maintain phase alignment between channels assessed by computing  $|s_{BF}^T|^2$

# Simulation in GNU Radio



$$Sum\ signal(\theta, \theta_{DOA}) = \sum_{n=1}^4 e^{j(n-1)\frac{2\pi}{\lambda} d(\sin\theta - \sin\theta_{DOA})}$$

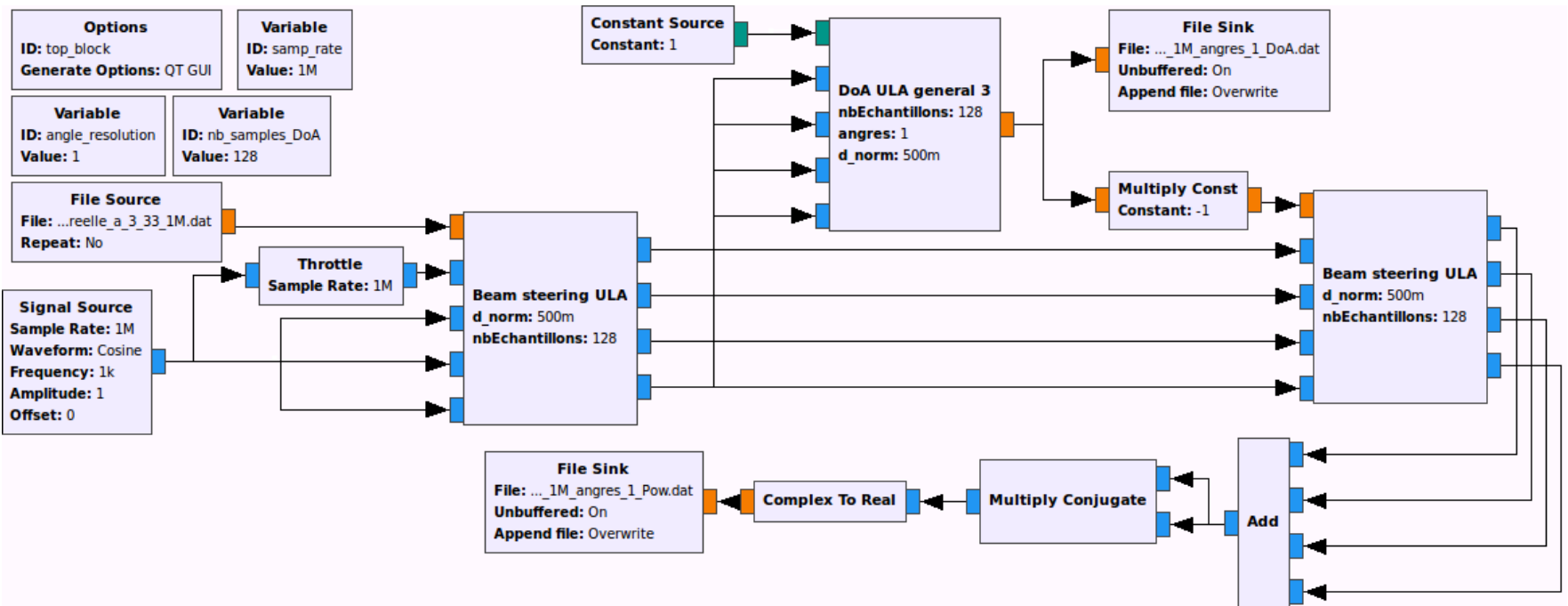
# Simulation in GNU Radio



$$\text{Sum signal}(\theta, \theta_{DOA}) = \sum_{n=1}^4 e^{j(n-1)\frac{2\pi}{\lambda} d(\sin\theta - \sin\theta_{DOA})}$$

$$G_{\text{reception simulation}}(\theta, \theta_{DOA}) = \left| \sum_{n=1}^4 e^{j(n-1)\frac{2\pi}{\lambda} d(\sin\theta - \sin\theta_{DOA})} \right|^2$$

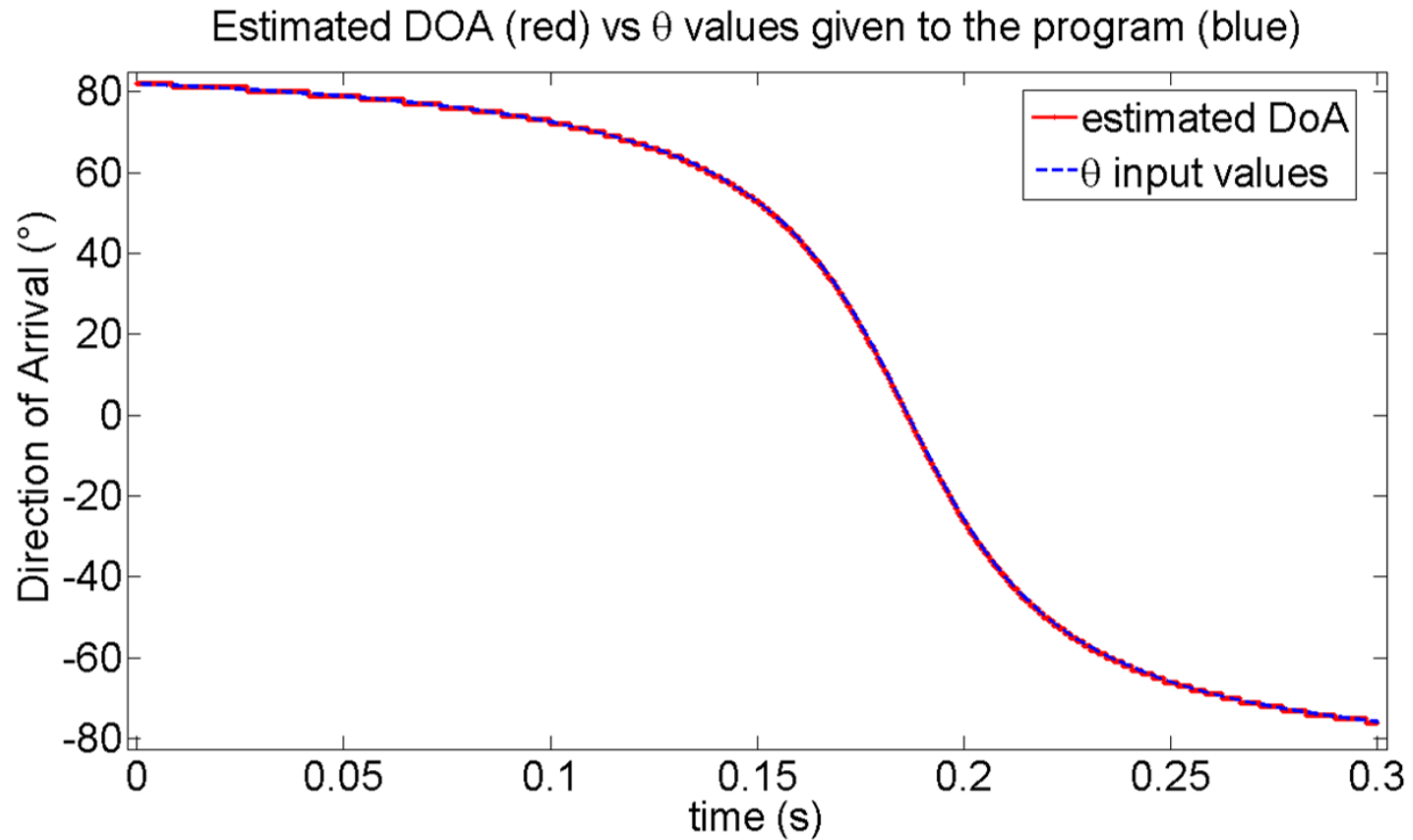
# Simulation in GNU Radio



Presented simulation is for a projectile flying at Mach 4,9 ( $\approx 1670$  m/s)

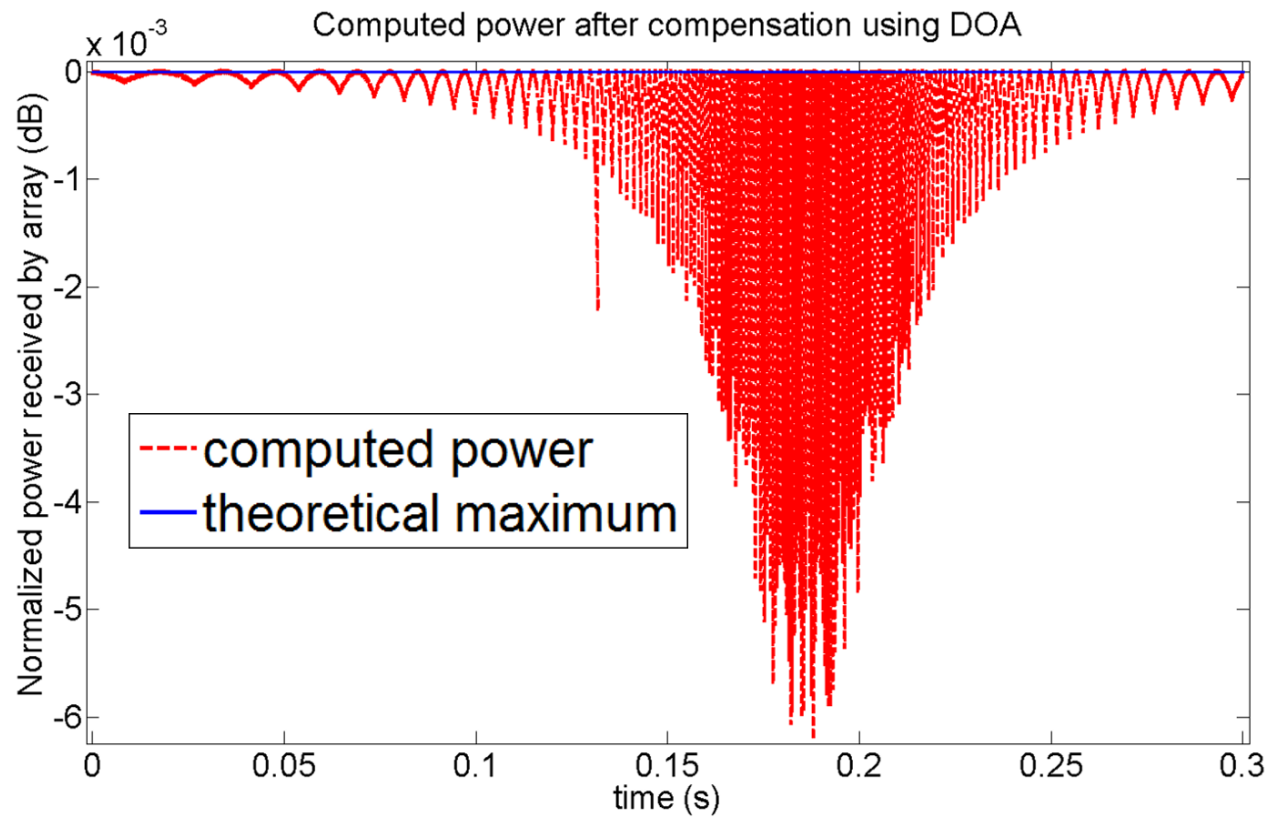
Sampling rate 1MS/s, DOA search  $1^\circ$  precise with 128 samples per DOA estimation

# Simulation in GNU Radio



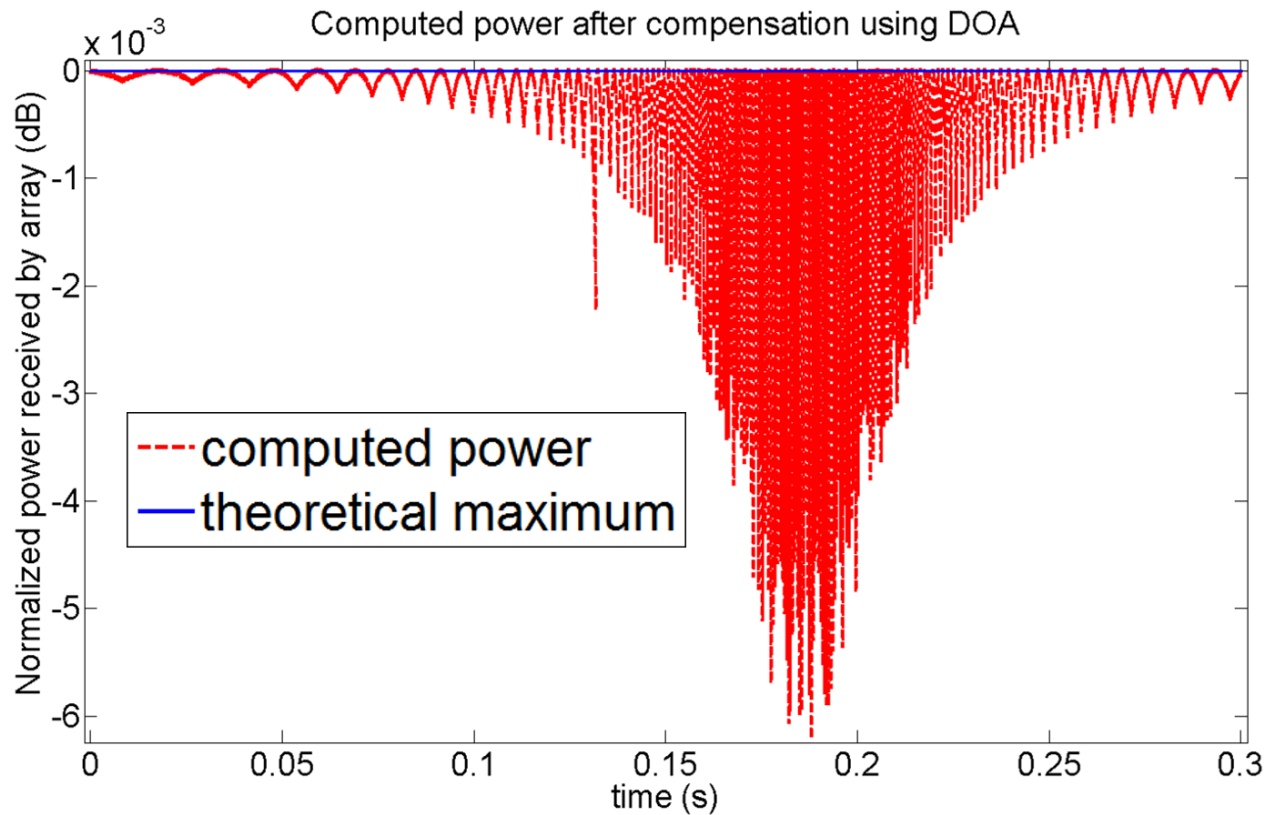
Simulation shows the program can accurately follow the projectile at Mach 4,9 (faster than real fired projectiles)

# Simulation in GNU Radio



Reception gain variations below  $10^{-2} dB$

# Simulation in GNU Radio



Reception gain variations below  $10^{-2} dB$



successfully maintained constant over projectile trajectory by accurately steering array main lobe in projectile direction.



# Overview

Background

Proposed solution

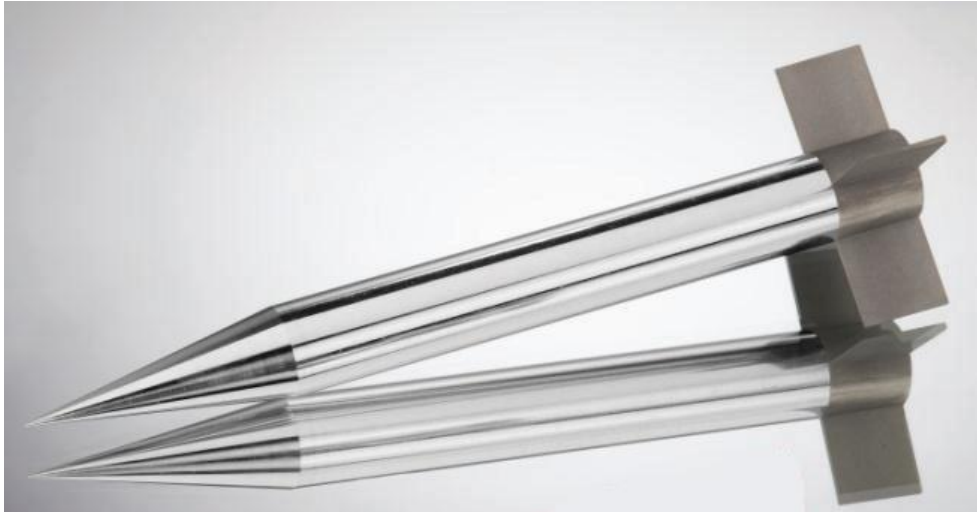
Used signal processing

Simulation in GNU Radio

**Measurement Results**

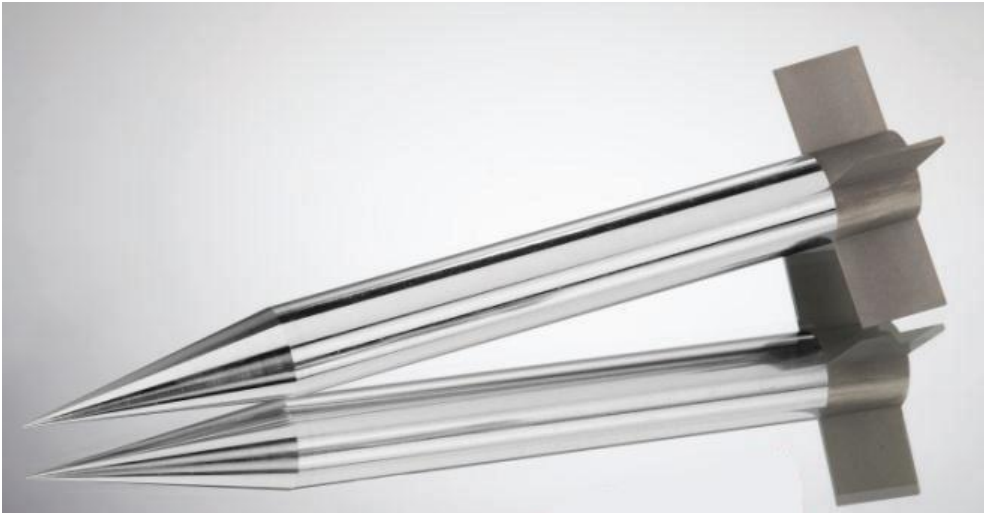
Conclusion and Outlooks

## Measurement results



Fired projectiles

# Measurement results

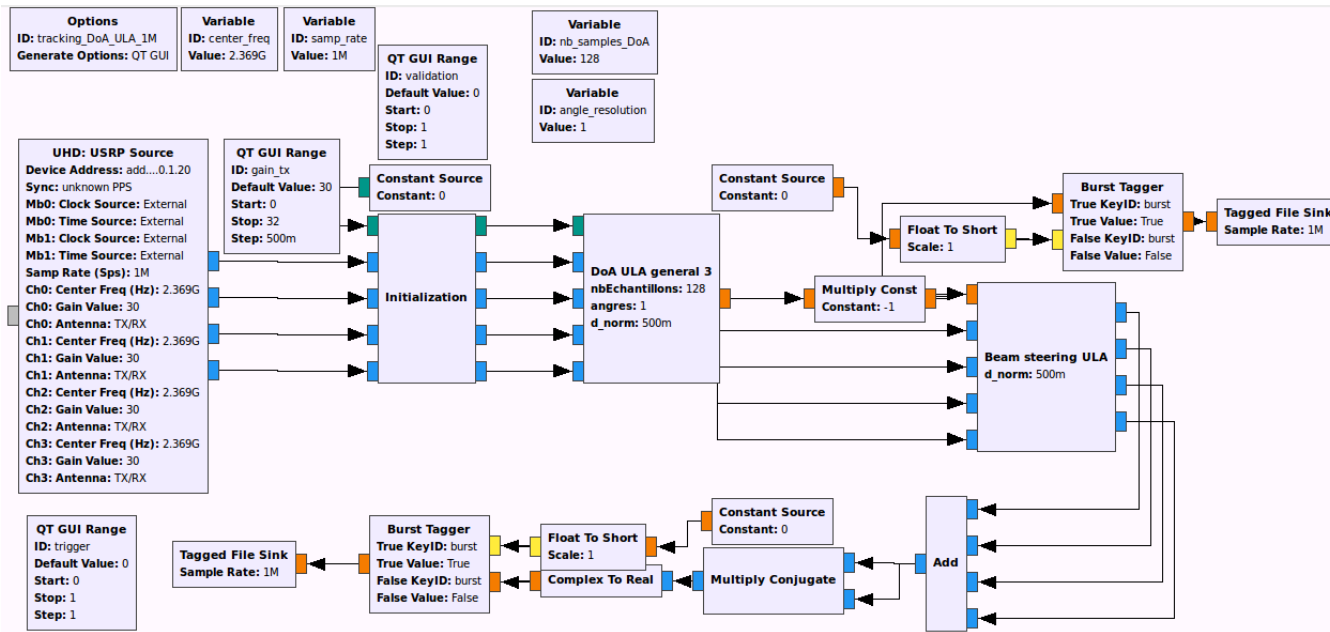


Fired projectiles



Embedded electronics

# Measurement results



Projectiles fired at Mach 1,3 in June

Transmitter onboard projectile GFSK 2Mbits/s @ 2,369 GHz

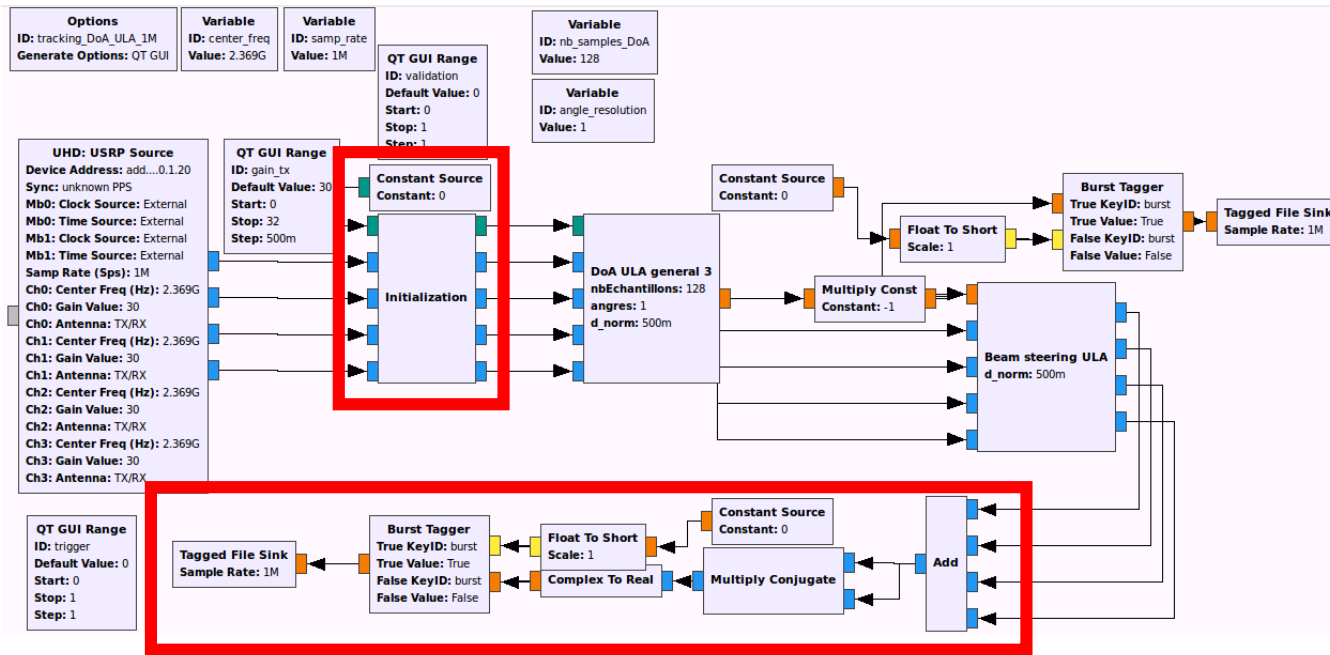
Projectile directly visible by antenna array in the [-60°; +50°] angular range

Conventional beamformer (Bartlett) algorithm

1MS/s sampling rate

128 samples per DOA estimation with 1° of angular resolution

# Measurement results

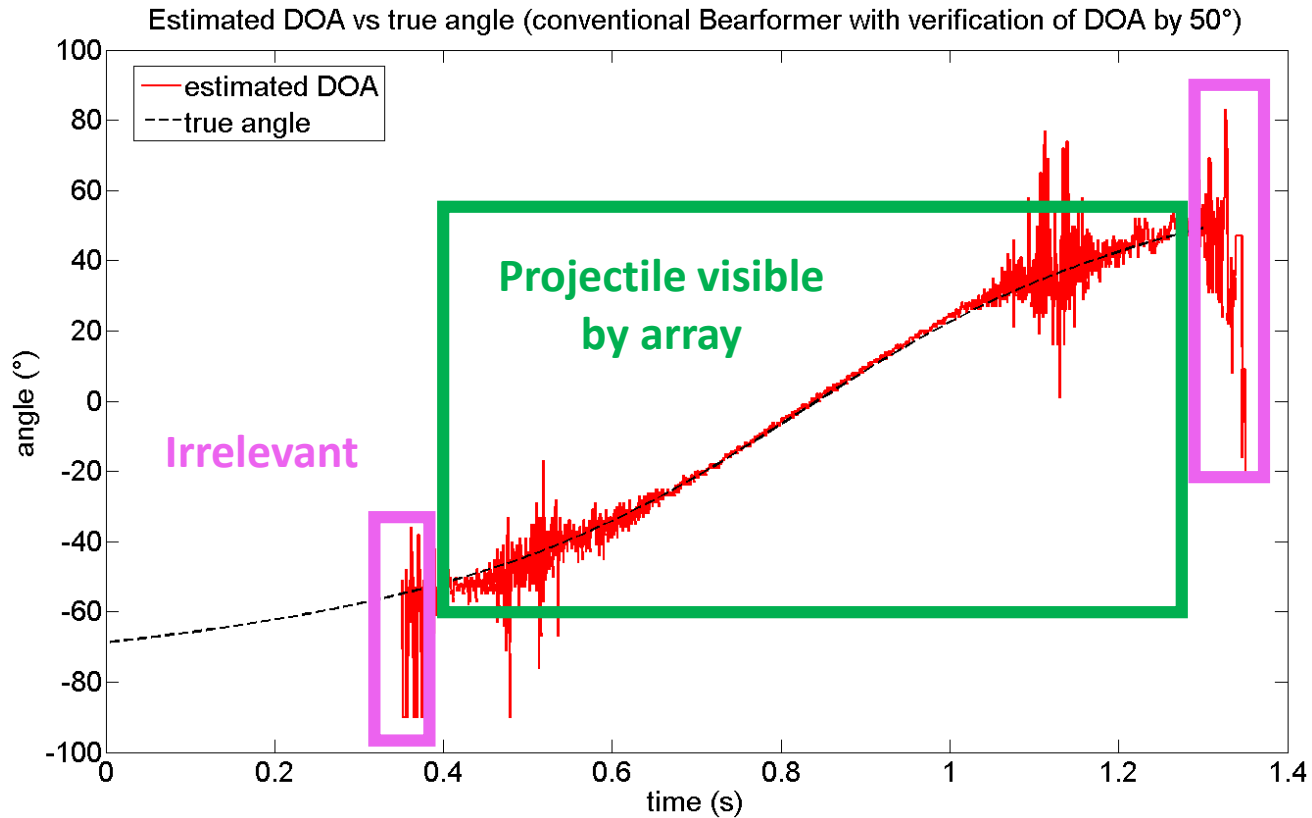


Compensation of phase-shifts created by hardware and antennas

$$G_{reception\ measurement}(\theta, \theta_{DOA}) = \left| \sum_{n=1}^4 e^{j(\varphi_{ant\ n}(\theta) - \varphi_{ant\ n}(\theta=0^\circ))} e^{j(n-1)\frac{2\pi}{\lambda}(\sin\theta - \sin\theta_{DOA})} \right|^2$$

# Measurement results

## 1<sup>st</sup> projectile

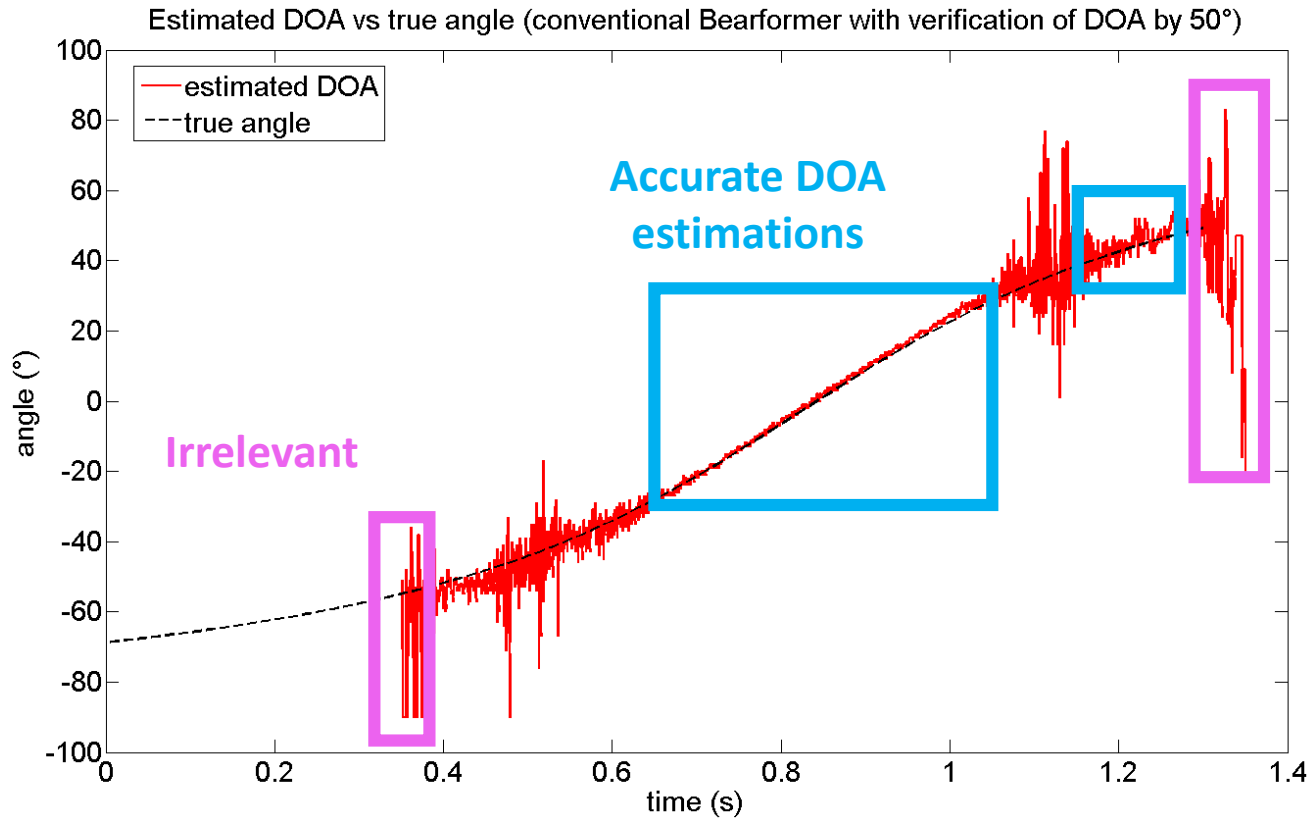


Projectile visible by array when in  $[-60^\circ; +50^\circ]$

Measurements outside  $[-60^\circ; +50^\circ]$  irrelevant

# Measurement results

1<sup>st</sup> projectile

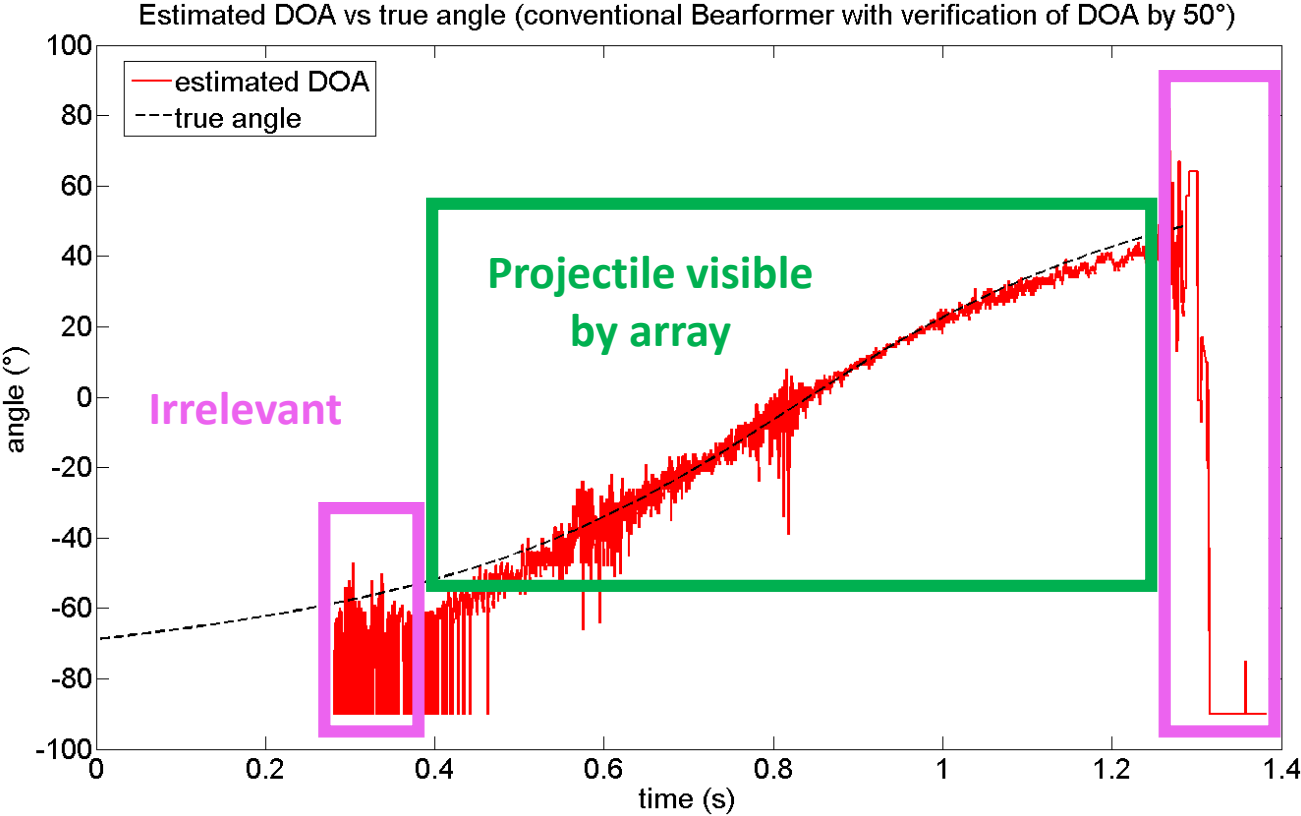


Accurate DOA acquisition in  $[-30^{\circ}; +30^{\circ}]$

Noise peaks between 0,5 and 0,6s, and between 1,1 and 1,2s.

# Measurement results

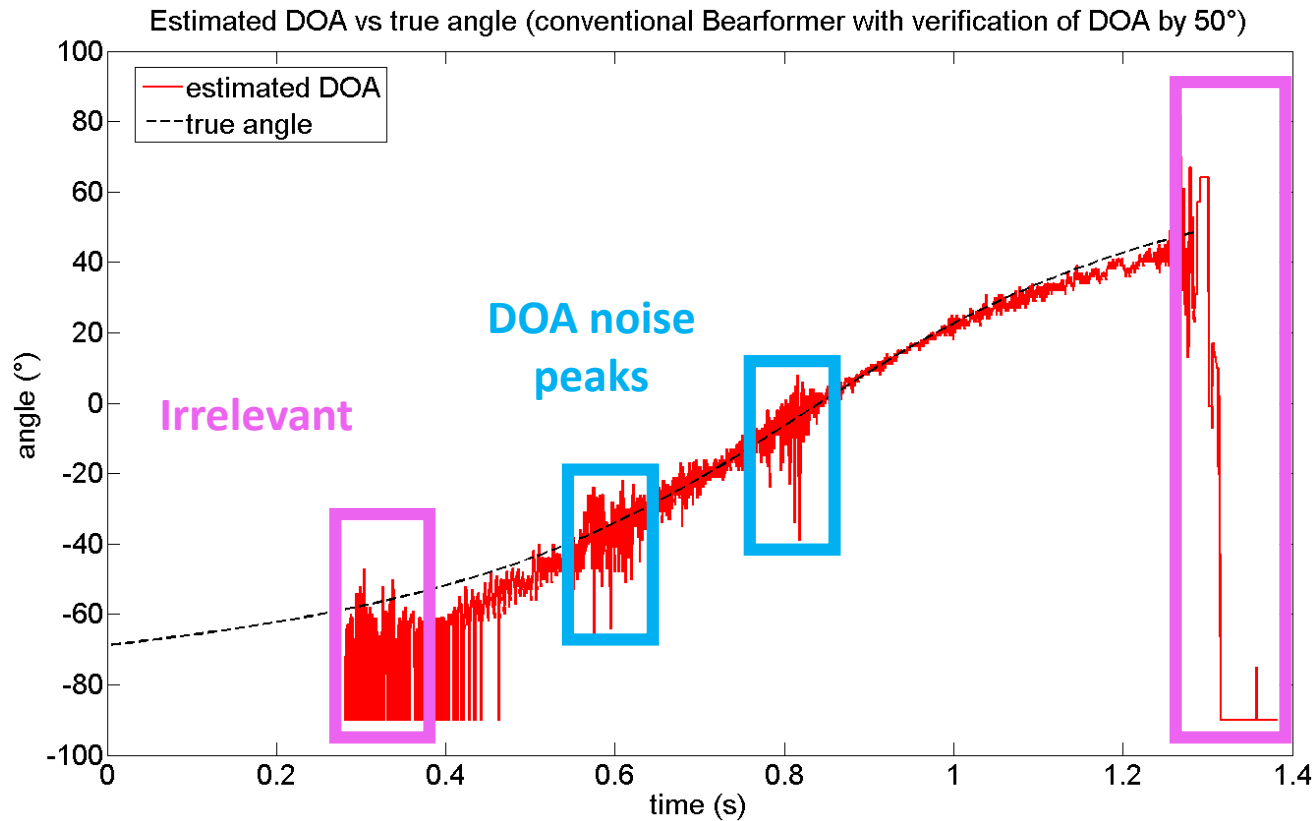
2<sup>nd</sup> projectile





# Measurement results

## 2<sup>nd</sup> projectile

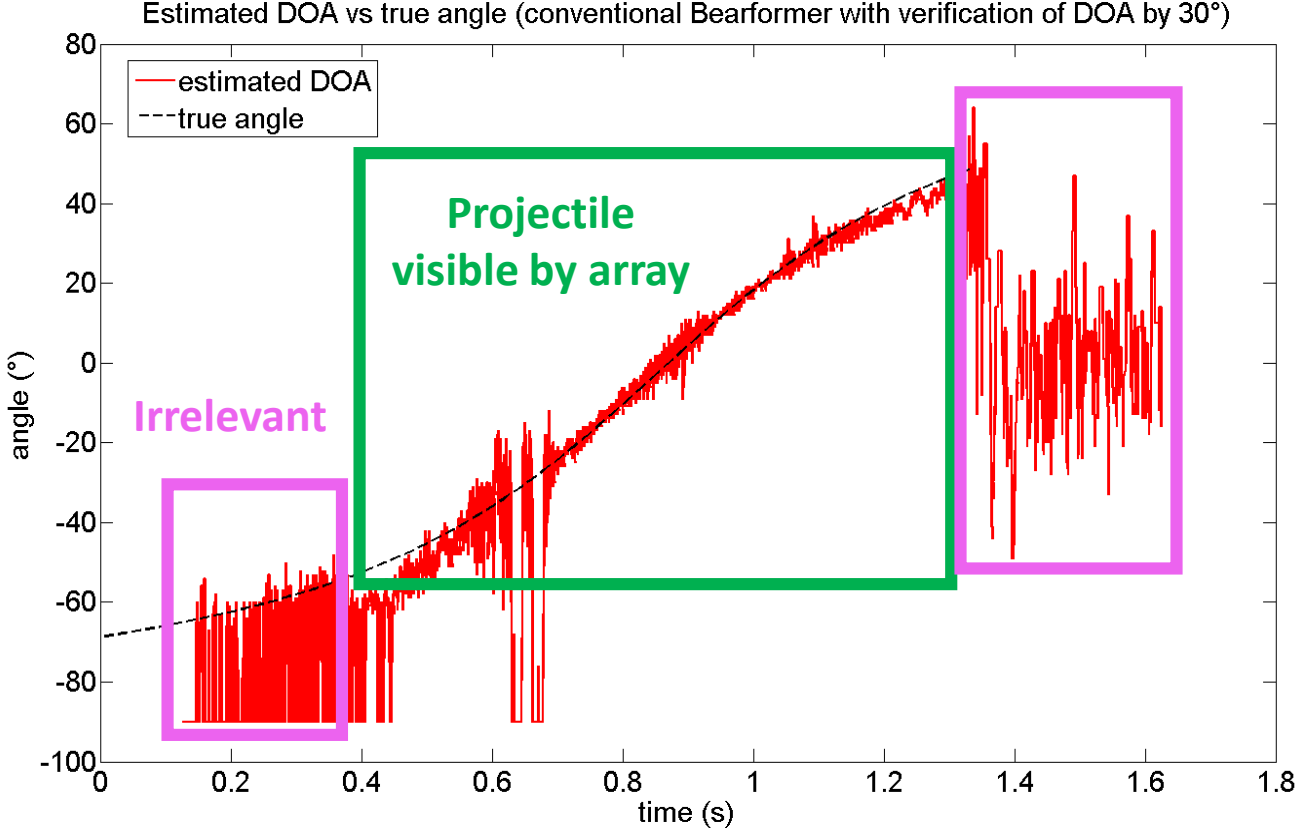


DOA acquisition noisier than with the 1<sup>st</sup> projectile

Noise peaks between 0,5 and 0,6s, and between 0,8 and 0,9s

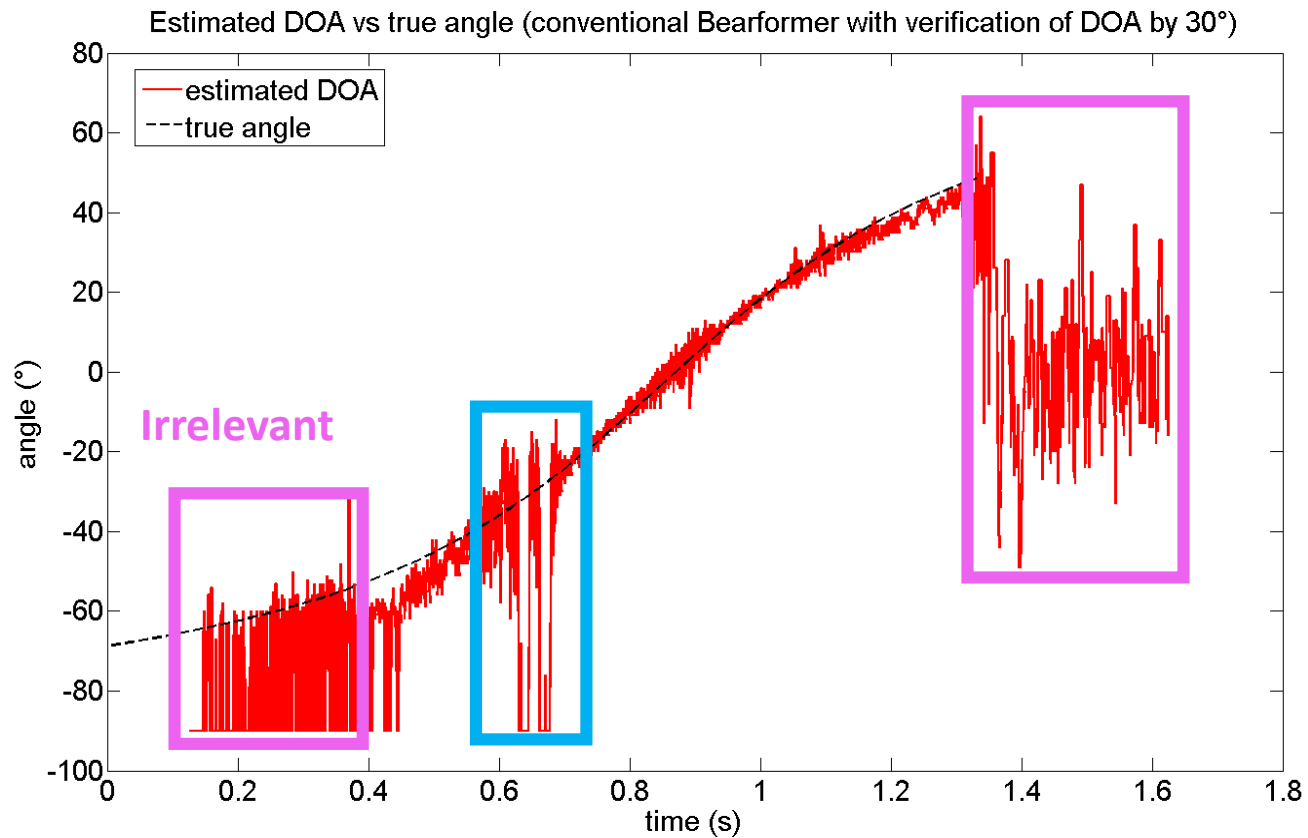
# Measurement results

3<sup>rd</sup> projectile



# Measurement results

## 3<sup>rd</sup> projectile

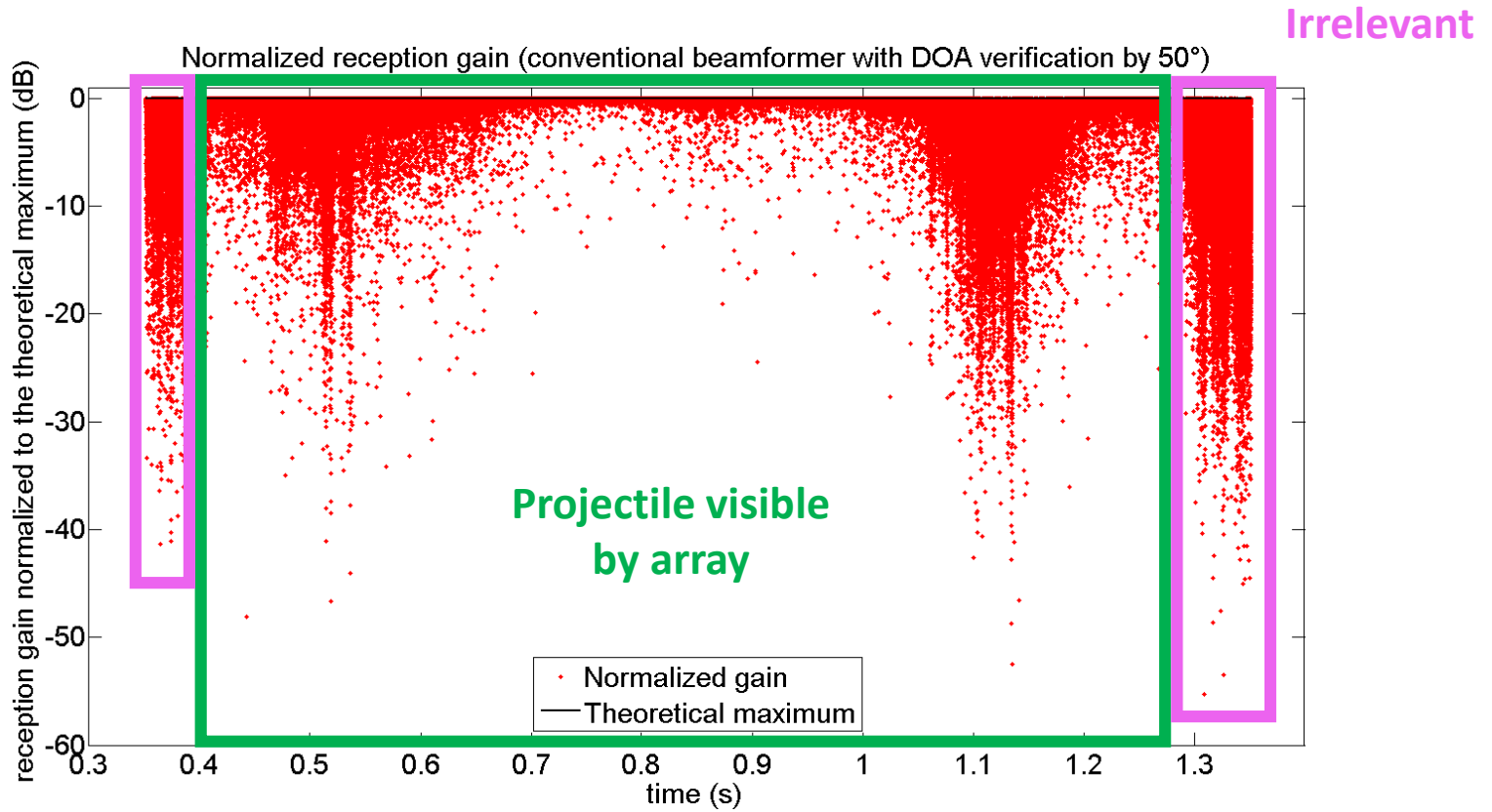


DOA verification loop blocked DOA at  $-90^\circ$  between 0,7 and 0,8s

➡ wrong estimates over 0,05s

# Measurement results

1<sup>st</sup> projectile

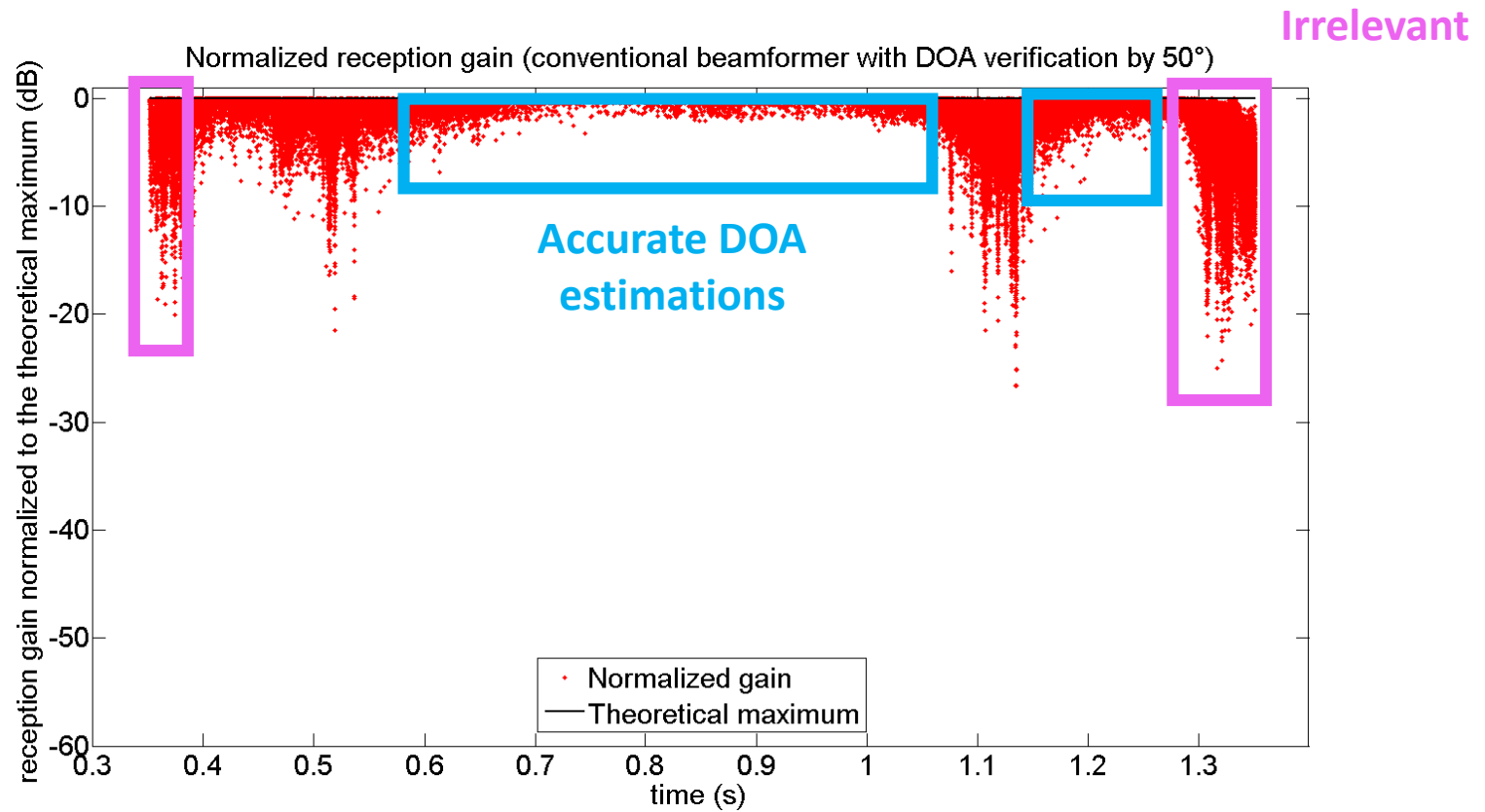


Some points indicate gain losses

# Measurement results

3 point averaging

1<sup>st</sup> projectile



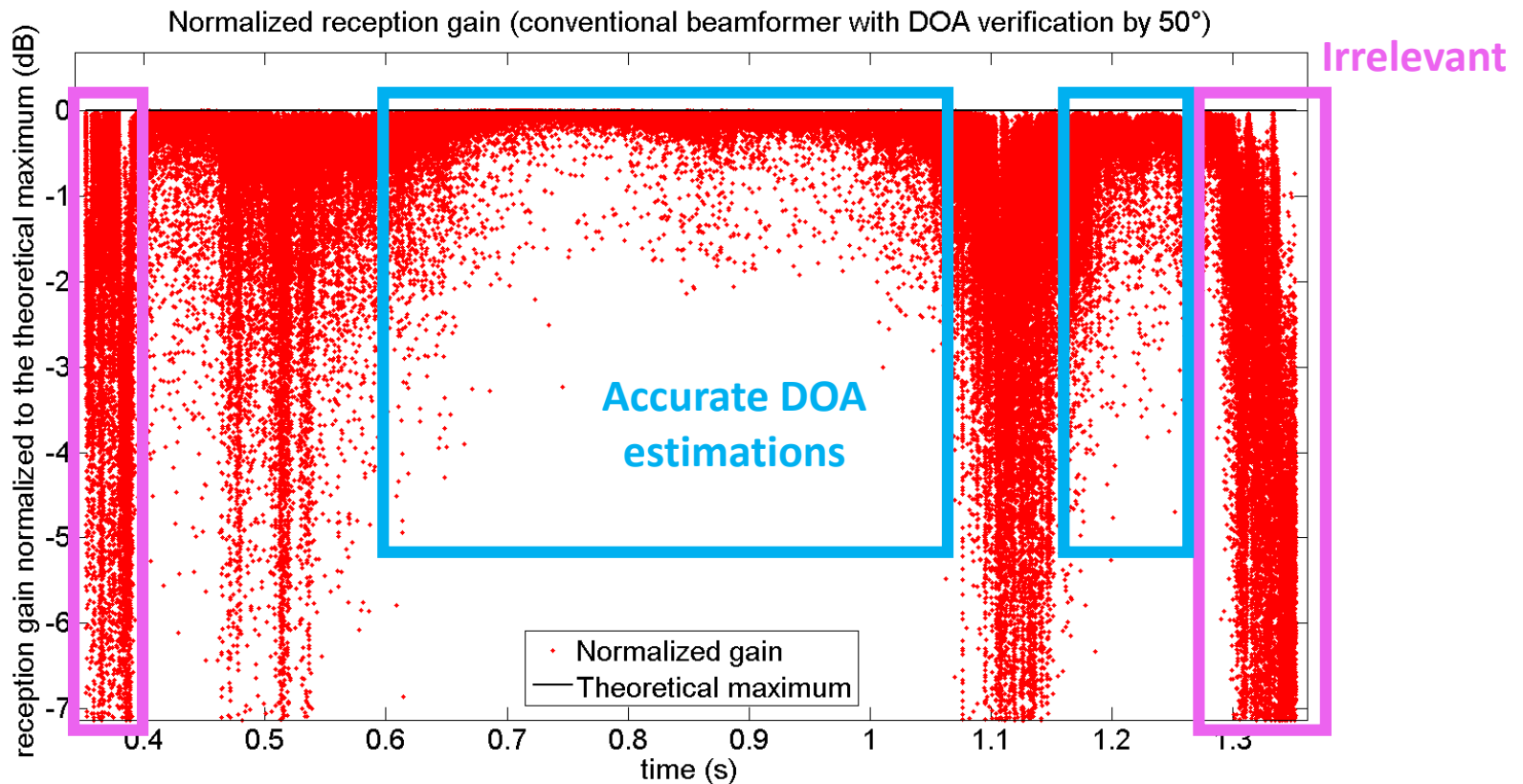
Gain losses represent a minority of points

Besides noise peaks, reception gain is constant within 5dB

# Measurement results

3 point averaging (Zoom)

1<sup>st</sup> projectile



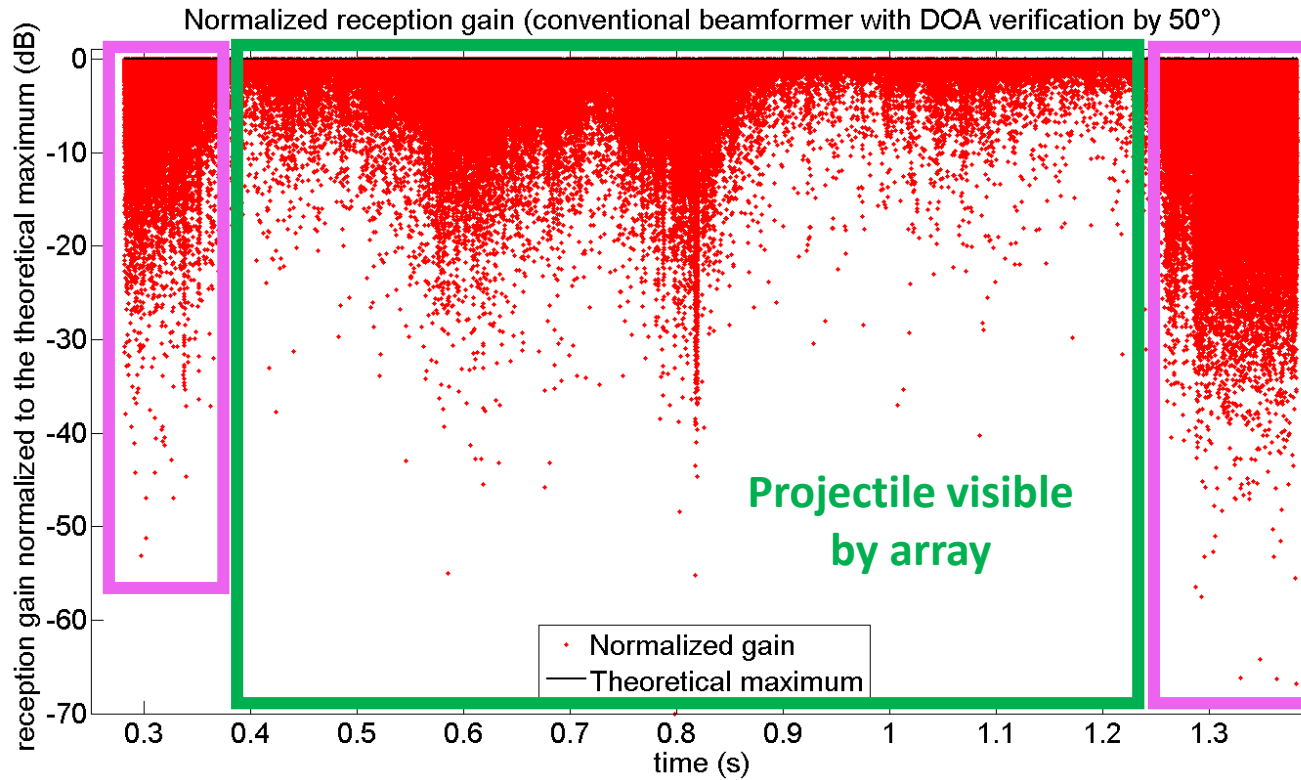
Gain losses represent a minority of points

Besides noise peaks, reception gain is constant within 5dB

# Measurement results

2<sup>nd</sup> projectile

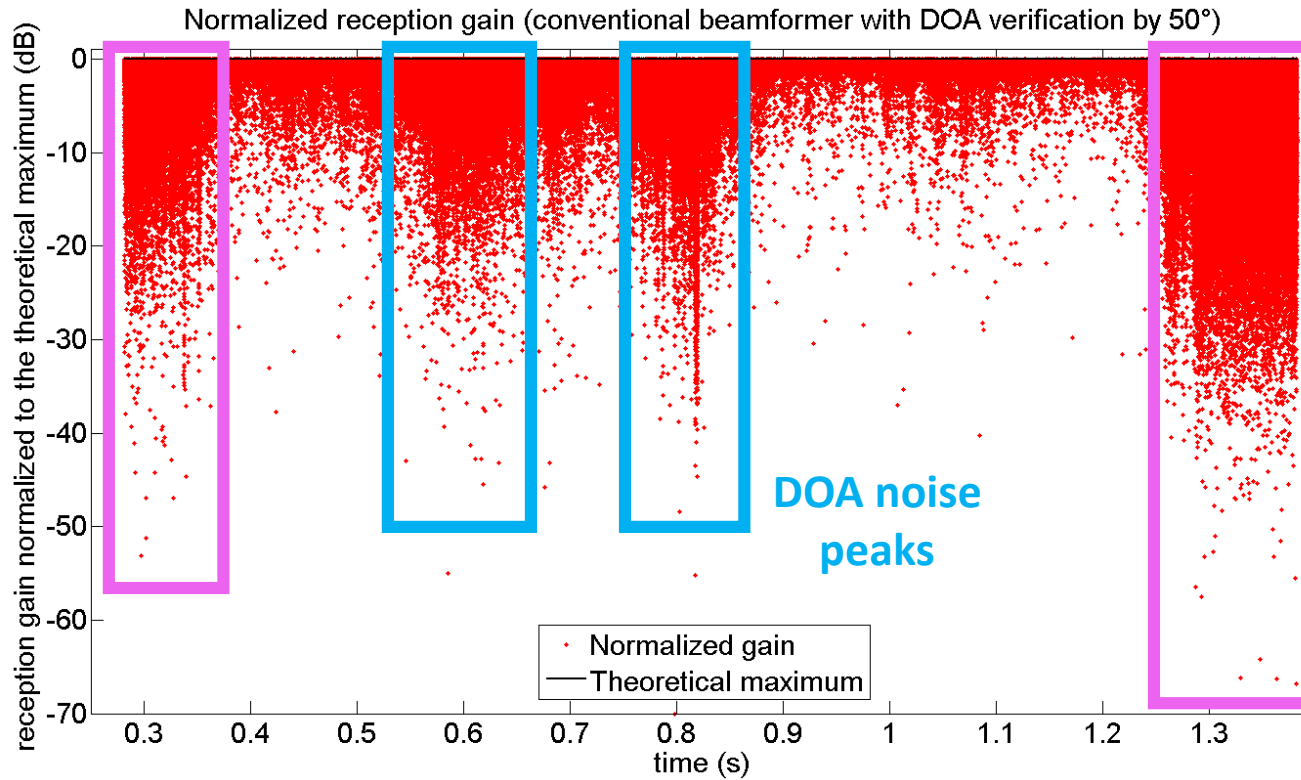
Irrelevant



Noise peaks on DOA estimations induce more gain losses

# Measurement results

2<sup>nd</sup> projectile



Noise peaks on DOA estimations induce more gain losses

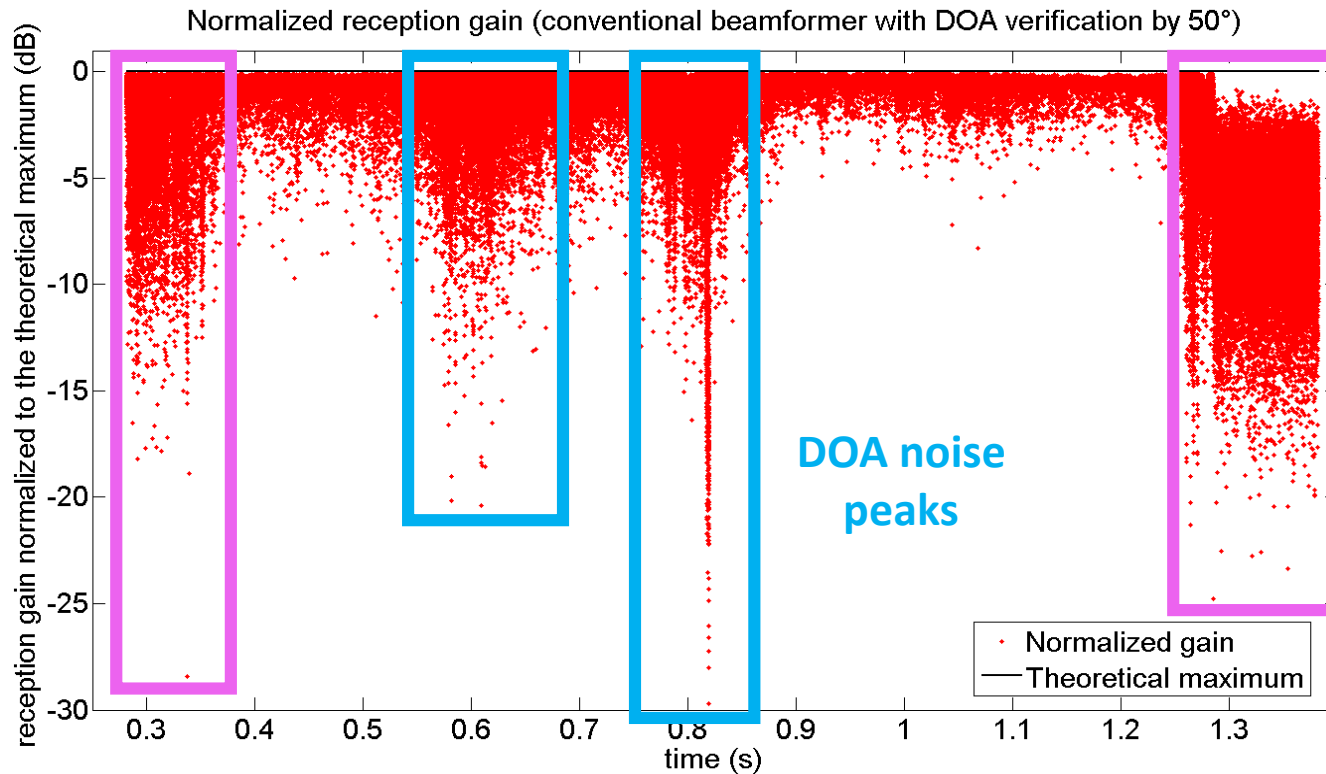


# Measurement results

2<sup>nd</sup> projectile

3 point averaging

Irrelevant

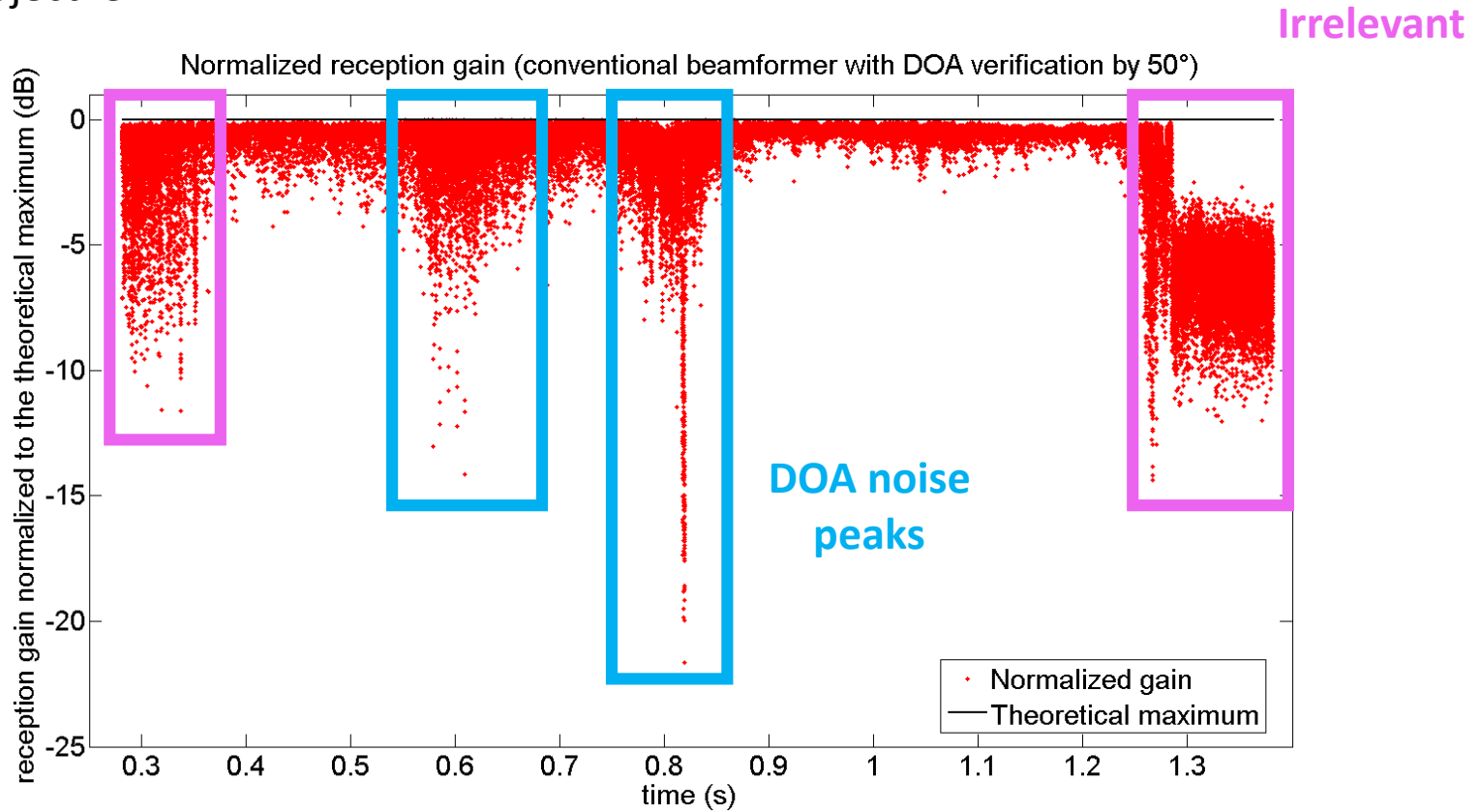


Majority of points remain above -5dB when correct DOA estimation

# Measurement results

2<sup>nd</sup> projectile

10 point averaging

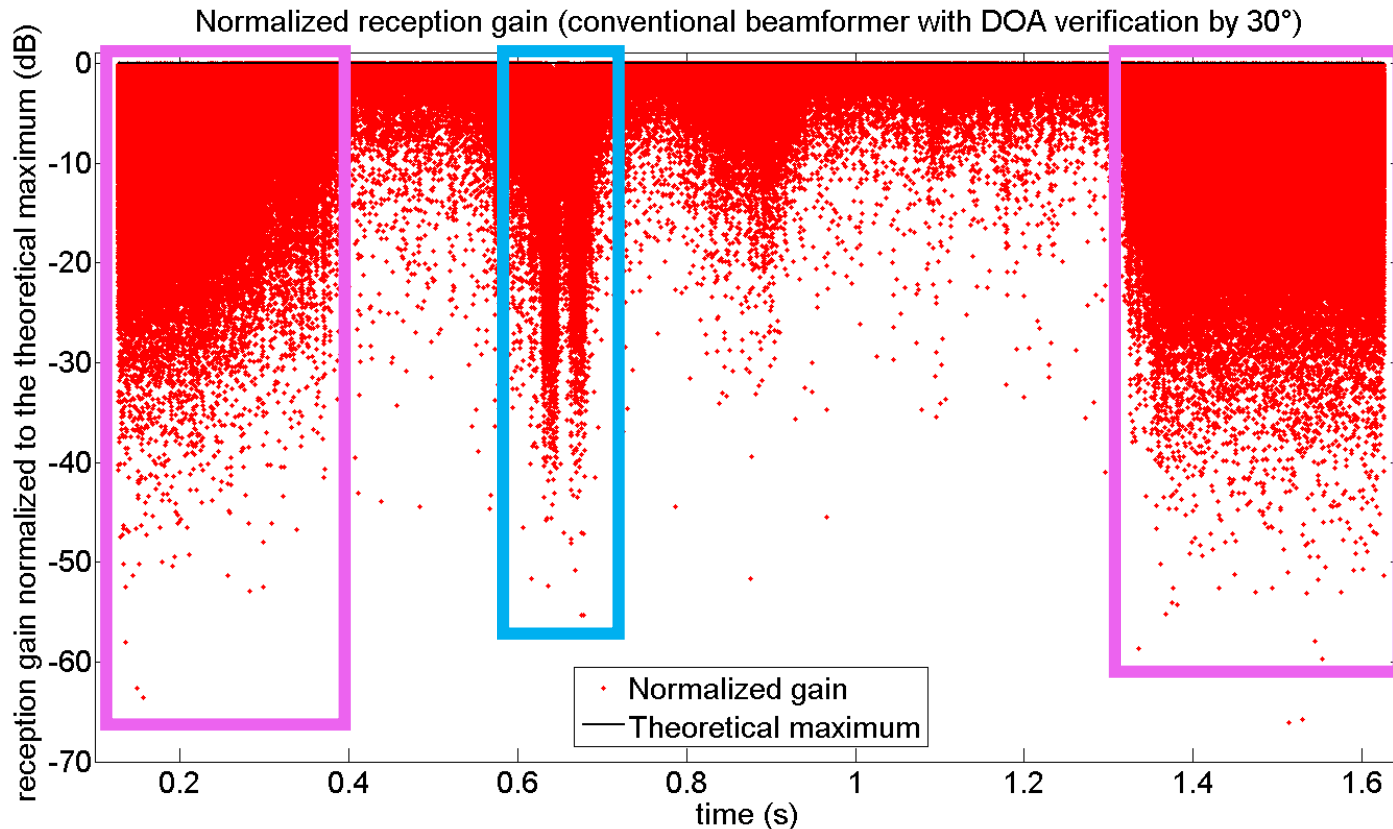


Majority of points remain above -5dB when correct DOA estimation

# Measurement results

3<sup>rd</sup> projectile

Irrelevant



DOA verification loop blocked DOA at  $-90^\circ$  between 0,7 and 0,8s

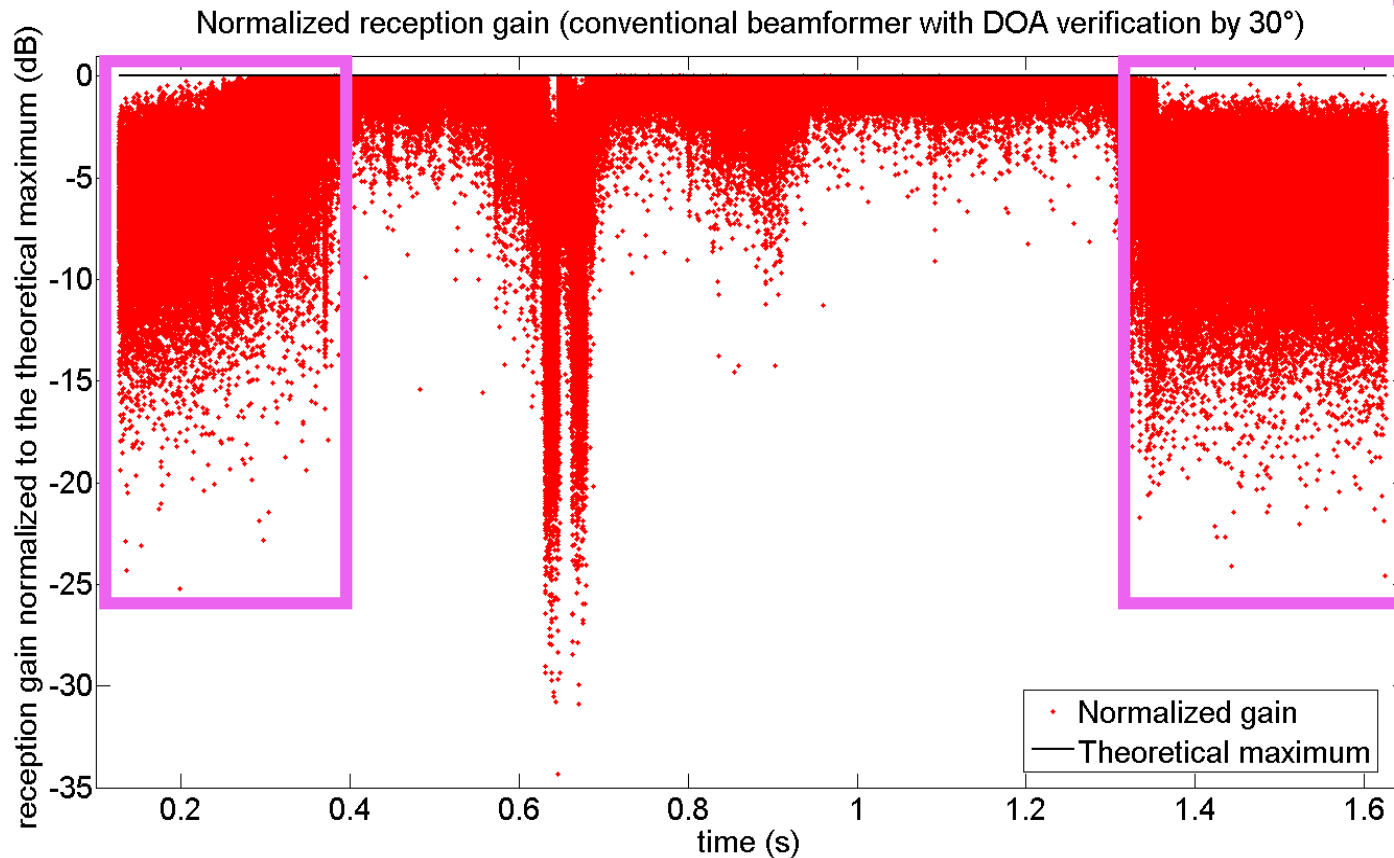
➡ wrong estimates over 0,05s

# Measurement results

3 point averaging

3<sup>rd</sup> projectile

Irrelevant



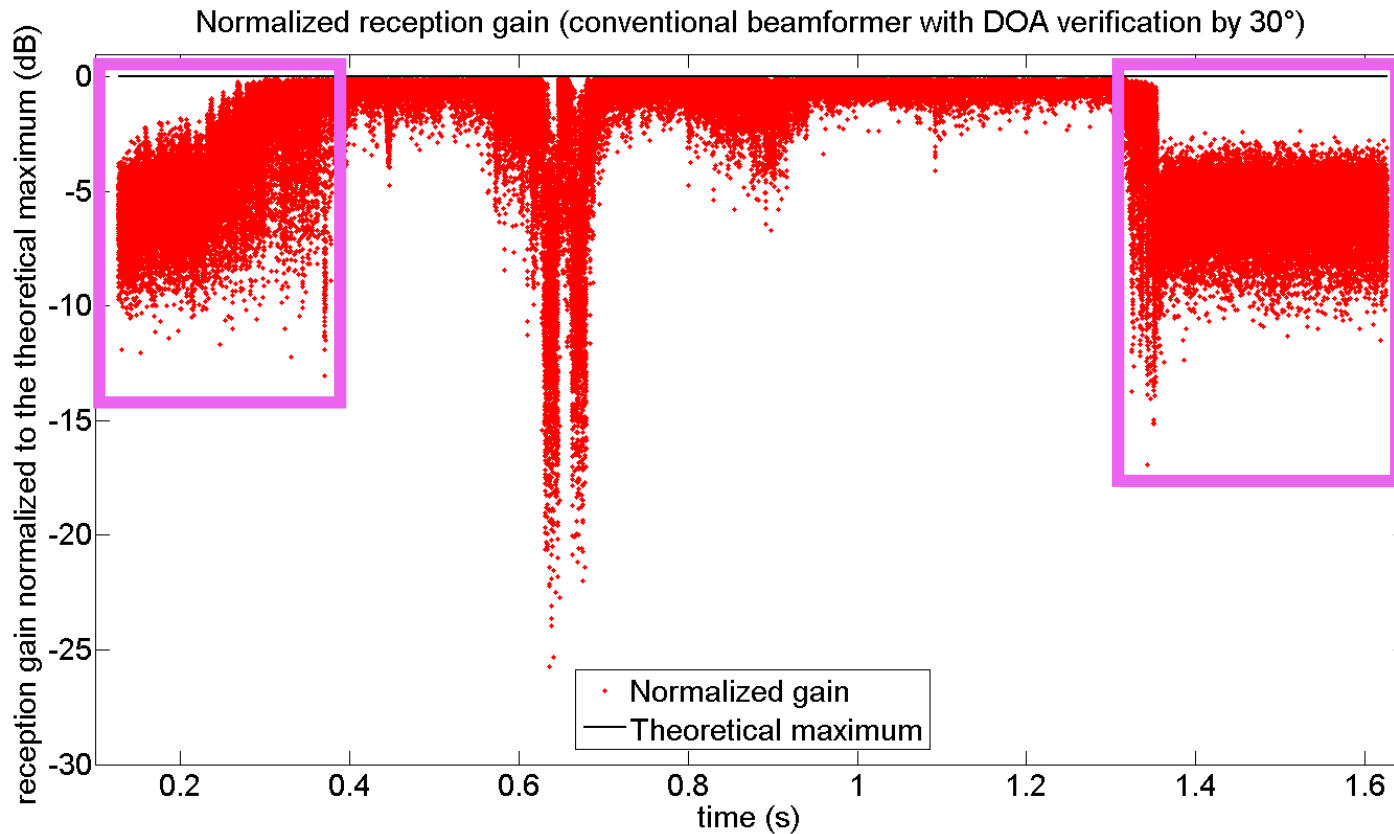
DOA errors aside, the averaged reception gain remains over -5dB over the projectile trajectory

# Measurement results

10 point averaging

3<sup>rd</sup> projectile

Irrelevant



DOA errors aside, the averaged reception gain remains over -5dB over the projectile trajectory

# Overview

Background

Proposed solution

Used signal processing

Simulation in GNU Radio

Measurement Results

**Conclusion and Outlooks**

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A passive solution for high-speed projectile tracking using commercial UBX-160 and GNU Radio is proposed

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First measurements have been performed with real projectiles and have demonstrated the system ability to follow projectiles in real conditions

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A passive solution for high-speed projectile tracking using commercial UBX-160 and GNU Radio is proposed

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Simulations run in GNU Radio show that the program and equipment are able to follow high-speed projectiles and gives a first insight on expected performance

First measurements have been performed with real projectiles and have demonstrated the system ability to follow projectiles in real conditions

Errors in DOA have been found, but the reception gain has been shown to remain constant within 5dB for the projectile trajectory with accurate estimation

# Outlooks

Measurements using other DOA algorithms (Capon)

# Outlooks

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Hardware update for remote measurement triggering

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Increase of the sampling rate to 25MS/s to run while performing telemetry reception measurements

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Research on embedding SDR components in gyrostabilized projectiles